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MICROPROCESSOR-BASED ON-LOAD VALVE
SEQUENCING FOR A TURBO-ALTERNATOR

by

M. R. OZGUR, B.Sc.

Thesis submitted for the Degree of Master
of Science in the Faculty of Science,
University of Durham.

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April, 1979.



ABSTRACT

An analog on-load valve sequencing system that is used to test the performances of the valves on a turbo-alternator can be replaced by a microcomputer testing system. On the way to the full-scale computerising of turbo-generator control systems, this type of testing system may be used under the control of a full-size supervisory computer. This microprocessor-based testing system provides test sequencing of the valves of an Electro-Hydraulic Governor. The hardware for the interfacing and scaling, and the firmware for the microprocessor are developed for one valve. The extension of the technique to the task of testing multiple valves in sequence is also discussed.

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C. A. Parsons & Co. Ltd., Newcastle Upon Tyne, is acknowledged for providing the author with technical knowledge of their governor model.

SUMMARY

Steam inlet valves on the turbine of^a turbine-alternator control the mechanical input power of the alternator. In turn, this input power effects the electrical output power of a generator. Fast valving control of the output power improves the system stability during the transient disturbances. For this reason, the inlet valves must operate efficiently and reliably.

In time, steam inlet valves get worn or overlapped because of the mechanical and chemical effects and their efficiency decreases. The valves are tested remotely from the central control rooms and the permanent records of the valves performance provided by^a X-Y plotter are collected, and the new records are checked against the previous ones. Comparison of the test traces with previous records highlights any change in valve performance. The records also indicate failure of valve operating systems, majority-voting circuits and valve tripping systems.

A typical Valve Governor System with its specifications was obtained from a manufacturer and reduced to be established on an analog computer. Then the mathematical model of the Governor on the analog computer was connected to a micro-processor over a designed hardware interfacing system.

The software was developed to do the same test steps^{as} done on actual valve. The test results are given. Some discussion is included on how the microprocessor-based on-Load valve testing technique can be extended to achieve the task of testing a number of valves.

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CHAPTER I

TURBO-GENERATORS

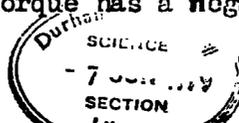
This chapter briefly describes some of the basic characteristics of synchronous machines and also provides the reader with an understanding of the methods of control of turbo-generators.

Attempts have been made to outline the essential points of a brief specification. The emphasis is on steam turbine driven generators and an outline analysis is given of such a machine.

1.1 Description and Torque Concept

In the synchronous machines field windings are supplied with direct current and alternating currents are obtained from or impressed on armature windings. Depending on the rotor shapes, synchronous machines are divided into two groups; salient pole machines and round-rotor machines.

Usually the armature windings are placed on the stators and the field windings on the rotors. Synchronous machines run at constant speed. The flux wave created by the armature current rotates in the air-gap at the same speed as the rotor rotates. So it looks stable as it is viewed from the rotor. The interaction between the magnetic field created by the field current and the magnetic field created by the armature current creates electro-magnetic torque. This torque is a function of the angle between the magnetic axis of the stator and the rotor or alternatively between the stator and rotor magnetomotive force waves (mmf) and opposes rotation in the generators. The electro-magnetic torque has a negative sign



for generator action where the negative sign indicates that the electro-magnetic torque acts in the direction to decrease the displacement angle between the field of stator and rotor, and it has a positive sign for motor action where the positive sign of the torque indicates that the electro-magnetic torque acts in the direction to bring these fields into alignment. A steady torque is produced if the displacement angle is kept constant which means that the stator and rotor magnetic fields travel at the same speed, as the magnetic fields of stator and rotor have constant amplitudes. [8]. Figure 1.1.

1.2 Voltage - Current Relations

The equations giving the four voltage-current relations in a three phase synchronous machine are

$$v_a = r i_a + \frac{d\psi_a}{dt} \quad (1-a)$$

$$v_b = r i_b + \frac{d\psi_b}{dt} \quad (1-b)$$

$$v_c = r i_c + \frac{d\psi_c}{dt} \quad (1-c)$$

$$v_f = r_f i_f + \frac{d\psi_f}{dt} \quad (1-d)$$

where v_s are the terminal voltages of the windings; i_a , i_b , i_c are the armature currents in the associated windings, i_f is the field current. r is the resistance of the armature winding and r_f is the resistance of the field winding. ψ is the flux linkage of each winding and can be

written as functions of currents, and the self and mutual inductances of each winding so that one can obtain four voltage equations as functions of these quantities. Since this gives a very complicated set of equations, a set of imagined voltages, currents and flux linkages are defined as functions of the actual variables. These new electrical variables called "Direct-axis and Quadrature-axis Variables" can be solved as functions of time and this solution can help to find the actual variables as functions of time. For a three phase synchronous machine, at the instant of Figure 1.2, the relation between the fictitious voltages and the actual voltages are as follows

$$\begin{bmatrix} v_d \\ v_q \\ v_c \end{bmatrix} = \frac{2}{3} \begin{bmatrix} \cos \theta & \cos (\theta - 120^\circ) & \cos (\theta + 120^\circ) \\ -\sin \theta & -\sin (\theta - 120^\circ) & -\sin (\theta + 120^\circ) \\ 1/2 & 1/2 & 1/2 \end{bmatrix} \begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} \quad (2)$$

$$\begin{bmatrix} v_a \\ v_b \\ v_c \end{bmatrix} = \begin{bmatrix} \cos \theta & -\sin \theta & 1 \\ \cos (\theta - 120^\circ) & -\sin (\theta - 120^\circ) & 1 \\ \cos (\theta + 120^\circ) & -\sin (\theta + 120^\circ) & 1 \end{bmatrix} \begin{bmatrix} v_d \\ v_q \\ v_c \end{bmatrix} \quad (3)$$

These equations can also be rewritten for the flux linkages and currents, by replacing \mathcal{V} with Ψ and i respectively. In the equations above, \mathcal{V}_d is the direct-axis voltage, \mathcal{V}_q is the quadrature-axis voltage and \mathcal{V}_0 is the zero-sequence voltage.

The self and mutual-inductances of a synchronous machine is the function of the displacement angle Θ between the direct axis and the phase axis, except the self-inductance of the field winding because the stator has a cylindrical shape.

Flux linkages as functions of currents can be obtained

$$\Psi_d = (L_s + M_s + \frac{3}{2} L_m) i_d + M_f i_f \quad (4-a)$$

$$\Psi_q = (L_s + M_s - \frac{3}{2} L_m) i_q \quad (4-b)$$

$$\Psi_0 = (L_s - 2M_s) i_0 \quad (4-c)$$

$$\Psi_f = \frac{3}{2} M_f i_d + L_{ff} i_f \quad (4-d)$$

where L_s is the component of the self-inductance of one phase of the armature which is independent of the angular

displacement θ , M_s is the component of the mutual inductance between two armature phases which is independent of the angle, M_f is the amplitude of the mutual inductance between the field winding and any armature phase, L_{ff} is the self-inductance of the field winding, L_m is the amplitude of the component of the stator self-inductance which is dependent on the angular displacement. Eqs.4 may be written

$$\psi_d = L_d i_d + M_f i_f \quad (5-a)$$

$$\psi_q = L_q i_q \quad (5-b)$$

$$\psi_o = L_o i_o \quad (5-c)$$

$$\psi_f = \frac{3}{2} M_f i_d + L_{ff} i_f \quad (5-d)$$

where $L_d = L_s + M_s + \frac{3}{2} L_m$, (6-a)

$$L_q = L_s + M_s - \frac{3}{2} L_m \quad (6-b)$$

and $L_o = L_s - 2M_s$ (6-c)

L_d , L_q and L_o are constant quantities and called direct-axis synchronous inductance, quadrature-axis synchronous inductance and zero-sequence inductance, respectively. These machine constants can be determined from tests.

From the equations given above, the following voltage-current relations are obtained

$$v_d = r i_d + L_d \frac{d i_d}{dt} - \omega L_q i_q + M_f \frac{d i_f}{dt} \quad (7-a)$$

$$v_q = r i_q + L_q \frac{d i_q}{dt} + \omega (L_d i_d + M_f i_f) \quad (7-b)$$

$$v_o = r i_o + L_o \frac{d i_o}{dt} \quad (7-c)$$

$$v_f = r_f i_f + L_{ff} \frac{d i_f}{dt} + \frac{3}{2} M_f \frac{d i_d}{dt} \quad (7-d)$$

These equations are for motor action. For generator action the currents i_d and i_q change the signs so that the equations become

$$v_d = -r i_d + \omega L_q i_q - L_d \frac{d i_d}{dt} + M_f \frac{d i_f}{dt} \quad (8-a)$$

$$v_q = -r i_q - L_q \frac{d i_q}{dt} + \omega M_f i_f - \omega L_d i_d \quad (8-b)$$

$$v_f = r_f i_f + L_{ff} \frac{di_f}{dt} - \frac{3}{2} M_f \frac{di_d}{dt} \quad (8-c)$$

[8,11] .

1.3 Power Equations and The Concept of Stability

In the steady state, the power output of a salient-pole generator is given as

$$P = E_q V \frac{r^2 + X_q^2}{r^2 + X_d X_q} \sin(\delta + \alpha) - V^2 \frac{r}{r^2 + X_d X_q} + V^2 \frac{X_d - X_q}{2(r^2 + X_d X_q)} \sin 2\delta \quad (9)$$

where E_q is the steady-state internal voltage of the generator, V is the terminal voltage, δ is the phase angle between the internal voltage E_q and the terminal voltage V , r is the resistance of the armature winding, $X_d = \omega L_d$ is called the direct-axis synchronous reactance and $X_q = \omega L_q$ is called the quadrature -axis synchronous reactance. (ω is the stator angular frequency and $\theta = \omega t$ where θ is the displacement angle). Here

$$\sin \alpha = \frac{r}{\sqrt{r^2 + X_q^2}} \quad (10)$$

and

$$\cos \alpha = \frac{X_q}{\sqrt{r^2 + X_q^2}} \quad (11)$$

If the machine is connected to the infinite bus V is the bus voltage, if it is connected to another machine V is the internal voltage of the second machine. If the connection was made through an internal impedance, r , x_d and x_q would be added to the external impedance when the equivalent circuit replacement of these impedances was being done. If the resistance is negligible, this power output equation of the generator simplifies to

$$P = \frac{E_q V}{X_d} \sin \delta + V^2 \frac{X_d - X_q}{2X_d X_q} \sin 2\delta \quad (12)$$

This equation was shown graphically in Figure 1.3.

If the machine had round rotor, x_d would be equal to x_q and therefore the reluctance power due to saliency represented by the second harmonic term would cease to exist, so that the power equation for the round rotor machine becomes

$$P = \frac{E_q V}{X_d} \sin \delta \quad (13)$$

In the case of the constant E_q and V , the maximum power occurs at $\delta = 90^\circ$ and this power limit is known as the Steady State Stability Limit.

The values of the voltages and the currents are rms values.

In the transient state, the power-angle expression becomes

$$P = \frac{E'_q V}{X'_d} \sin \delta + V^2 \frac{X'_d - X_q}{2 X'_d X_q} \sin 2\delta \quad (14)$$

The voltage E'_q is the quadrature-axis voltage behind transient reactance, and X'_d is the direct-axis transient reactance. The second harmonic is negative since usually $X_q > X'_d$. The amplitude of the transient power-angle curve is greater than that of the steady state. So, if the duration of the suddenly applied overload is short, synchronism will not be lost. Refer to Figure 1.4.

For a synchronous machine the accelerating power is given as

$$P_a = P_i - P_o = I \omega \frac{d^2 \delta}{dt^2} \quad (15)$$

where P_i = Input power of the machine

P_o = Output power of the machine

I = Moment of inertia

ω = Angular velocity

and

δ = Angular displacement of the machine with respect to another machine or to the infinite bus to which it is connected.

At steady state the accelerating power equals zero and the power input equals the power output. Negative P_a represents retarding power and positive P_a represents accelerating power. The P_a versus δ curve shows that the system is unstable if the angle δ increases without limit. For generator action P_1 is the mechanical input power and P_e is the electrical output power. As derived earlier, this electrical output power is represented by the sinusoid on the power-angle curve.

The equal-area criterion is used to determine the Transient Stability Limit. As the assumptions of constant input power, constant voltage behind transient reactance and no damping are made, the generator and the infinite bus are in balance at point a (Figure 1.5) on the pre-fault power curve. P_1 is constant and therefore is parallel to the horizontal axis. The power-angle curve during the fault is lower than the pre-fault curve. When a fault occurs electrical power output decreases and the operating point a drops to the point b on the fault output curve. At this point mechanical input power exceeds electrical output power. This accelerates the unit causing the angle δ increase. The acceleration causes the electrical output power to increase to the point c. Since at point c the accelerating power decreases but the speed of the machine is still high the angle increase to δ_m . This results in retarding power on the generator. Therefore the speed decreases as does the angle, and the operating point moves back towards c. The generator will not pull out of step

during the transient disturbance if the area A_1 equals or is less than the area A_2 on the diagram. This is the definition of the transient stability limit. If the area A_1 is greater than the area A_2 , the generator stays in synchronism with the system.

1.4 The Voltage Regulator

One of the ways of improving the stability margin is by increasing the internal voltages of the synchronous machine.

Since the power output of a synchronous machine is proportional to its excitation voltage this output power is controlled by varying the field current. This variation may be achieved by use of an Automatic Voltage Regulator (AVR) or by manual control. By increasing the excitation voltage the amplitude of the transmitted power equation increases. When the steady-state limiting angle exceeds 90° a fast acting AVR increases the applied voltage to the field circuit of the machine; in other words the demagnetising effect of the armature current is being opposed by maintaining the decaying field flux linkage by means of AVR.

1.5 Fast Turbine Valve Control

The technique of fast turbine valve control has been introduced elsewhere by some authors [1]. The principle idea of this technique is to enhance generating-unit stability for

transient disturbances by increasing the critical fault clearing time.

Soon after a fault recognition the controller is designed to close the turbine valves for a short time period related to the time-integral of the difference between input power and output power, and then re-opened slowly. This action of the valves results in a decrease in the mechanical input power. By this means, the difference between the reduced generator electrical output power and mechanical input power is reduced quickly.

On referring to the power-angle curve in Figure 1.6, it can be seen that if a fault occurs while the turbine generator is in balance with the power system at point P, the electrical output power of the generator will drop to the point P1. After the fault clearance, the system is stable at newly established working point P2 on the post-fault power-angle curve. In the diagram of Figure 1.6, the mechanical input power is shown as a function of δ which arises from the fast valve operation. It is indicated by the dotted line. Fast valve action starts as the fault occurs.

Since the transient stability limit is given with $A_1 \geq A_2$, as the fast valve control concept is applied it is apparent that a significantly increased critical fault clearing time arises. [1].

1.6 Steam Turbine Governor Operation

The decreased inertia-to-torque ratio on large turbo-generator units increases the control complexity of turbines

so that electro-hydraulic control systems essentially replaced the conventional mechanical hydraulic governing systems.

In basic form, an electro-hydraulic governor is illustrated in Figure 1.7 [9]. Three main feedback loops are shown dotted. To this figure, a main stop valve just ahead of the main governing valves and a reheater stop valve just to the reheater output may be added [4].

There are two different modes of governor operation; "Full-Arc Admission" and "Partial-Arc Admission". As a brief explanation; in "Full-Arc Admission" operation, the control of steam flow to turbine is accomplished by throttling the main steam through main stop valve while the governor valves are fully open. In the later mode, however, the main steam is throttled through control valves while ^{the} stop valve is wide open. Full-Arc Admission operation is done during wide range speed control and Partial-Arc Admission operation is done during normal operation.

In conventional mechanical-hydraulic governor systems the speed and the load reference signals were both applied before the regulating element as shown in Figure 1.8. The block 'Valve Operators of Steam Volumes' represents all valve controlling loops in one generator control system.

An attempt to overcome the difficulty that arises during the transfer of the operation modes, from one mode to the other, has been implemented in electro-hydraulic control systems by making the load reference signal independent of regulation [4]. Refer to Figure 1.9.

Another proposed approach uses a conditional loop so that the selected higher loop error assumes the control [9].
Figure 1.10.

For the purpose of generator protection against overspeed there are usually two independent emergency governors provided. The one of them is Normal Speed Governor that operates Control Valves and Interceptor Valves, and the other is an emergency Trip system that shuts the Main Stop Valve and the Reheater Stop Valve on overspeed conditions.

Since the input power of a turbo-generator is controlled by means of steam inlet valves, stability can be markedly altered by their action. They must operate efficiently and reliably.

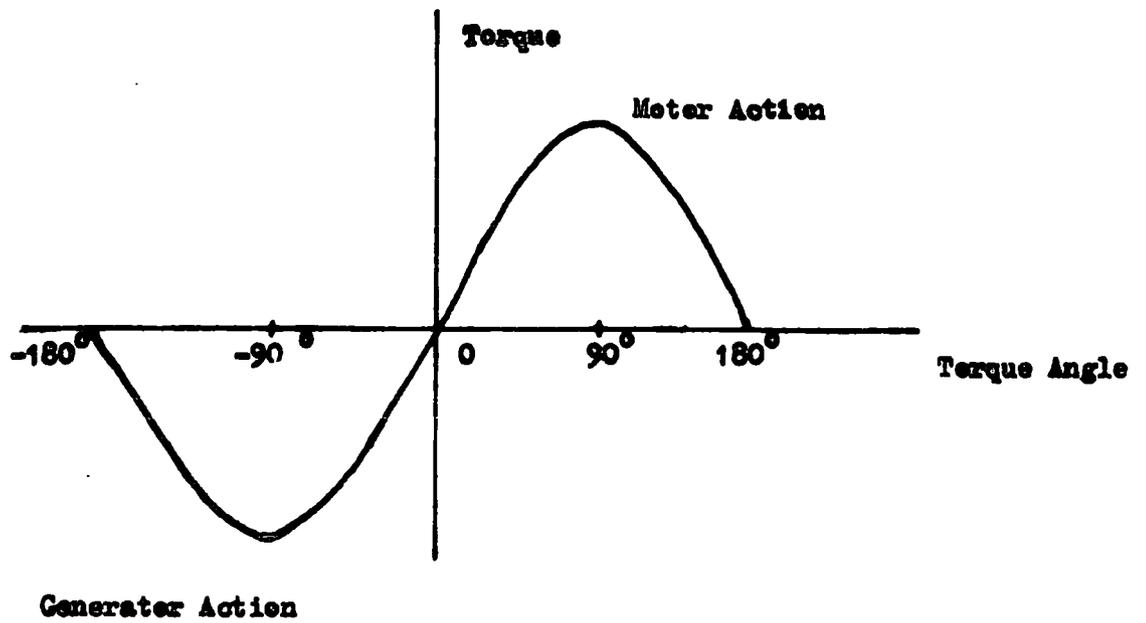


Figure 1.1 Torque-Angle Curve of a Synchronous Machine

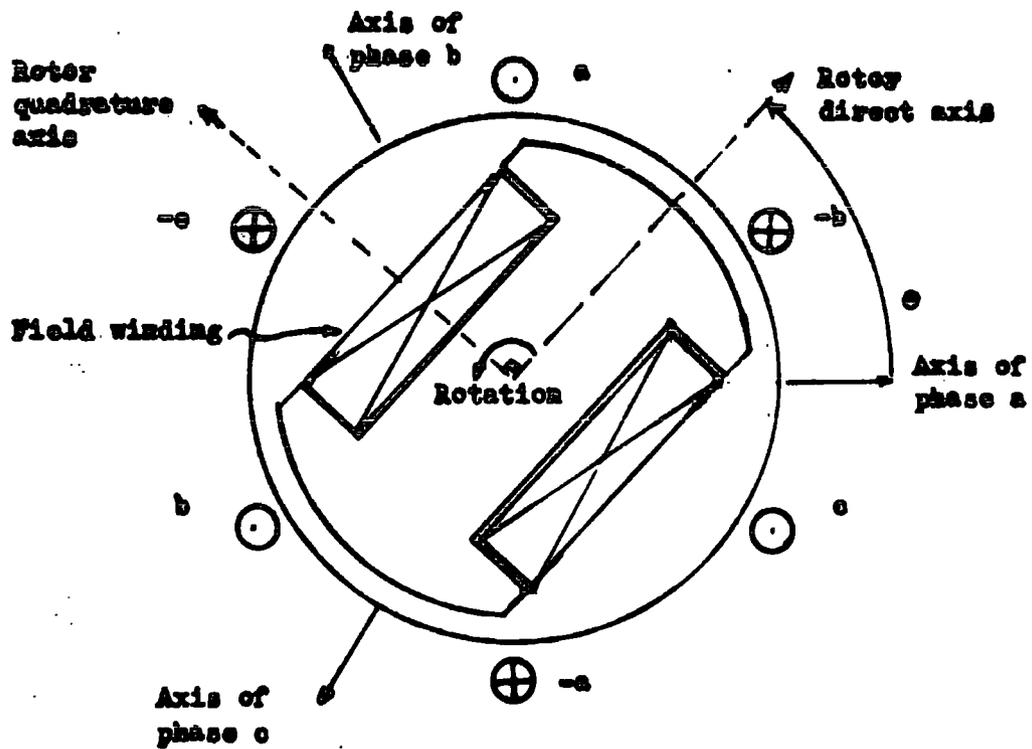


Figure 1.2 Elementary salient-pole 3-phase synchronous machine

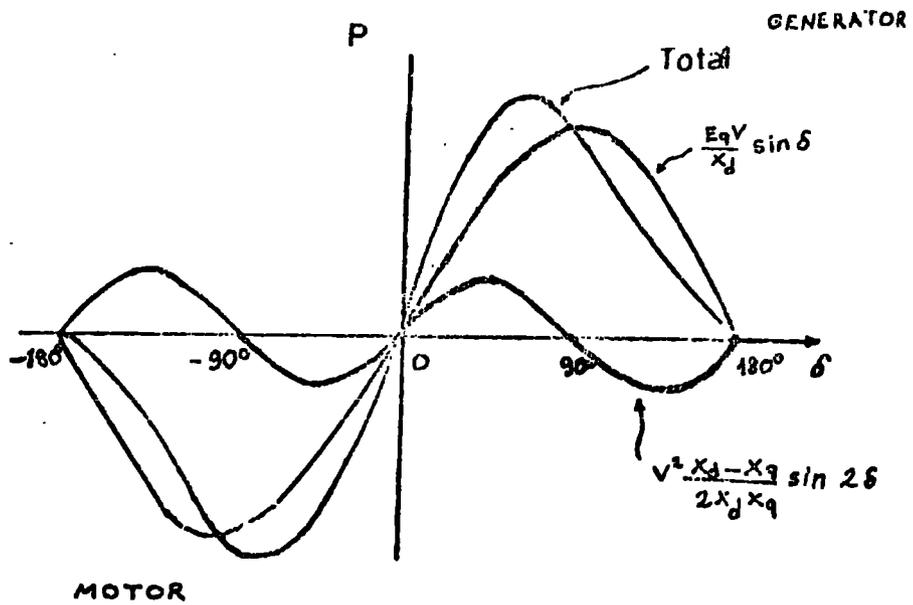


Figure 1.3 Steady-state power-angle curve of salient-pole synchronous machine.

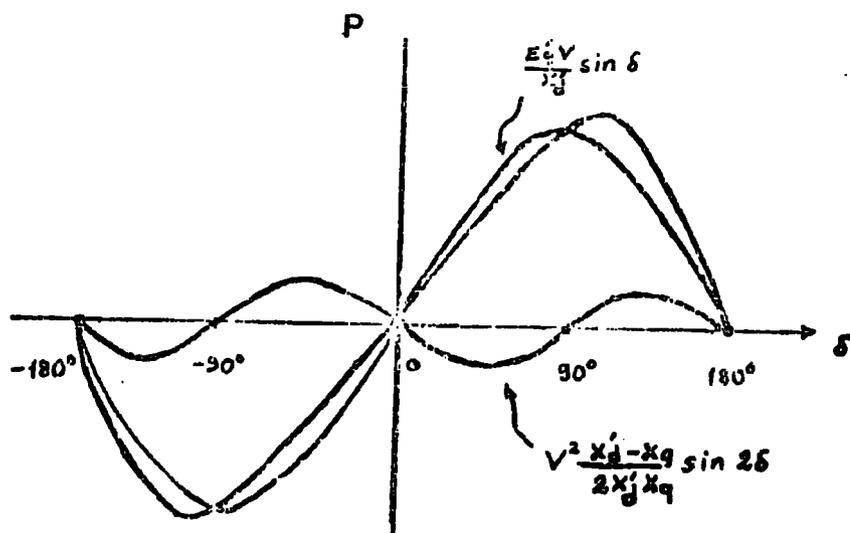


Figure 1.4 Transient power-angle curve of salient-pole synchronous machine.

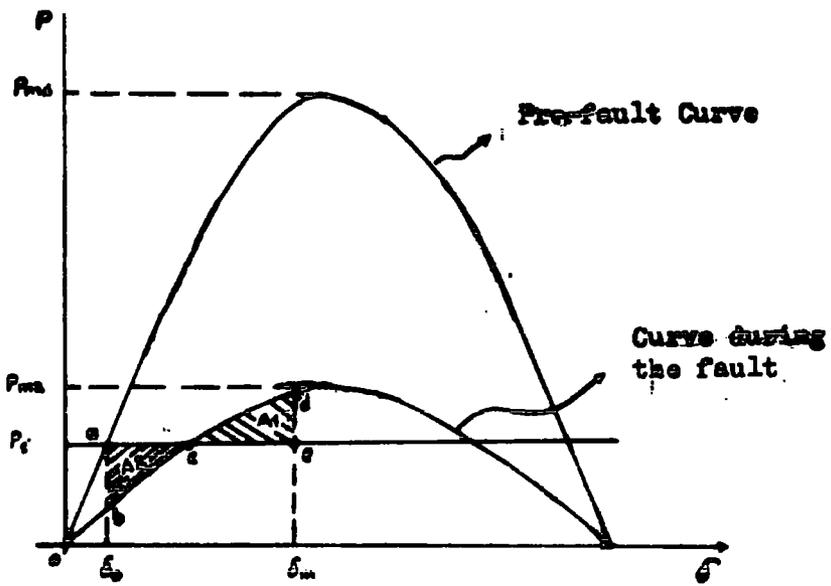


Figure 1.5 Determination of stability limit by the equal-area criterion.

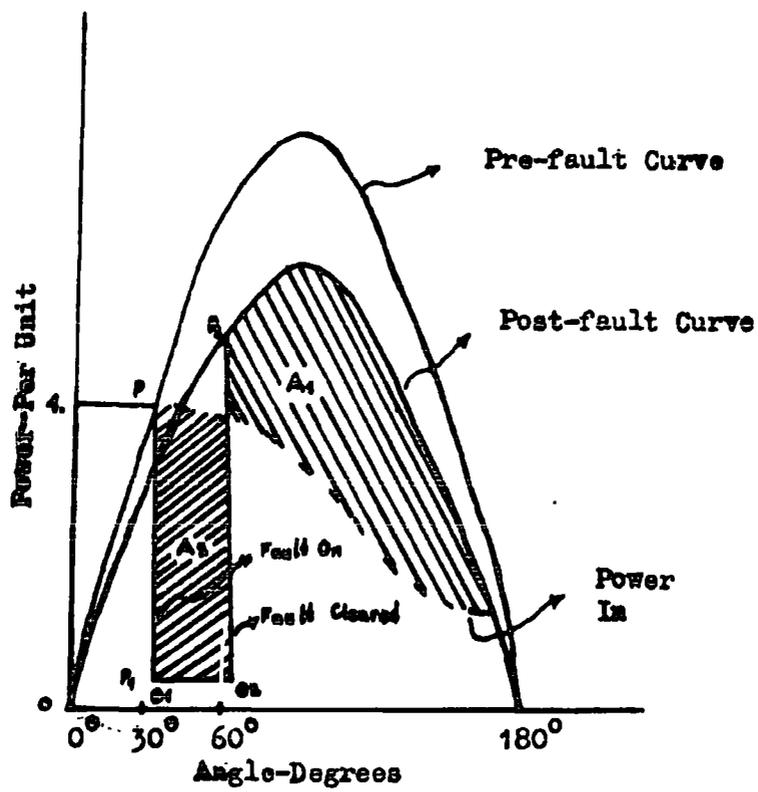


Figure 1.6 Transient-stability improvement by fast valve action.

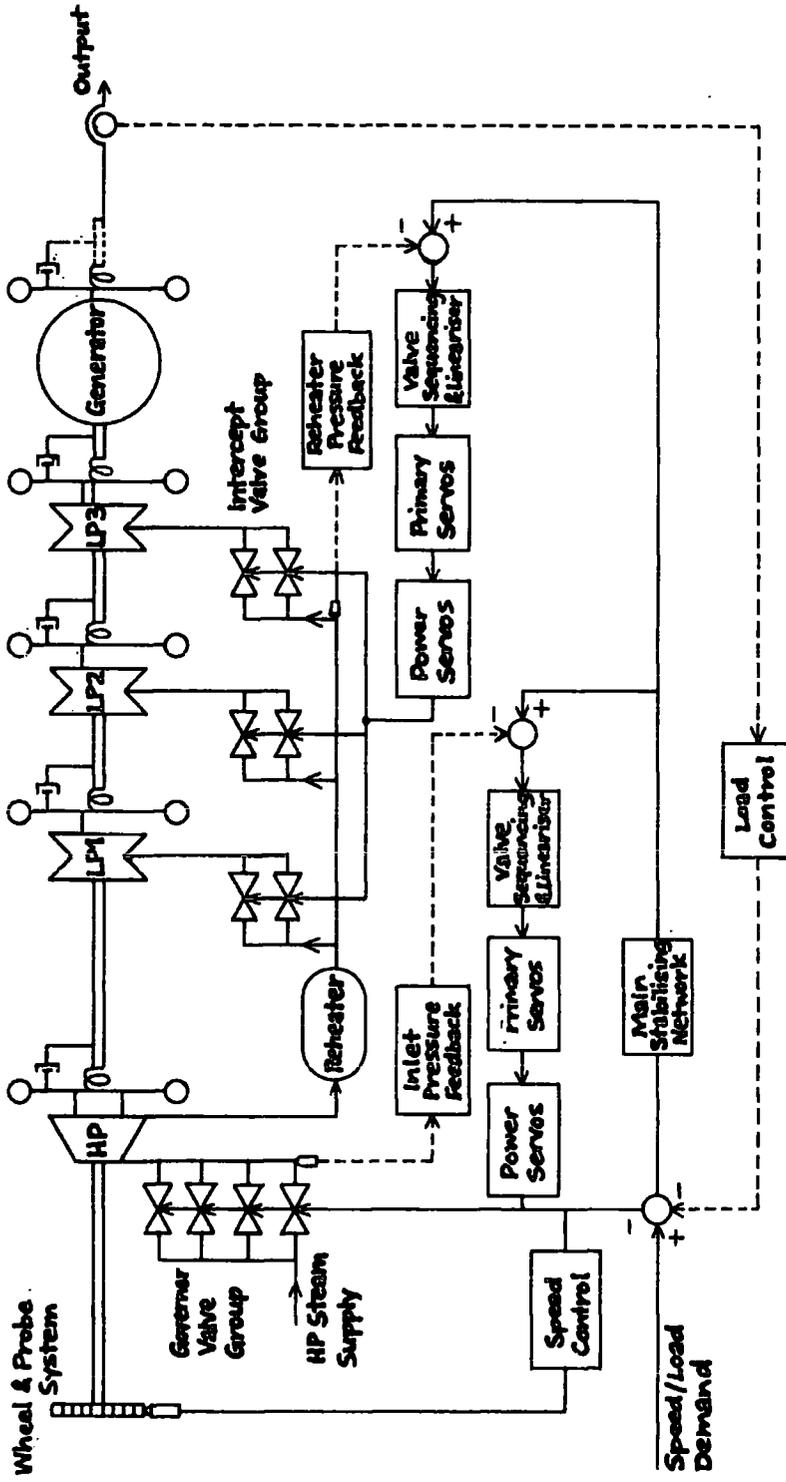


Figure 1.7 A typical Electro-Hydraulic Governor system block diagram of a Turbo-Generator

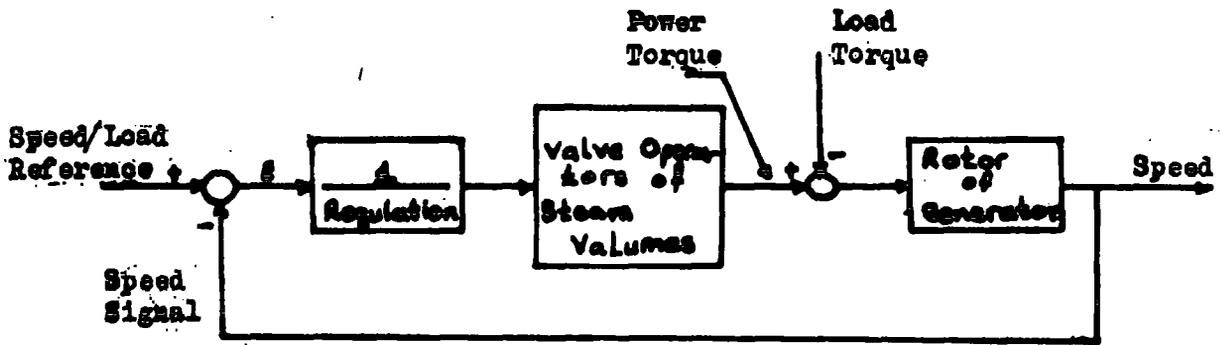


Figure 1.8 Block diagram of speed control loop for conventional turbine control system.

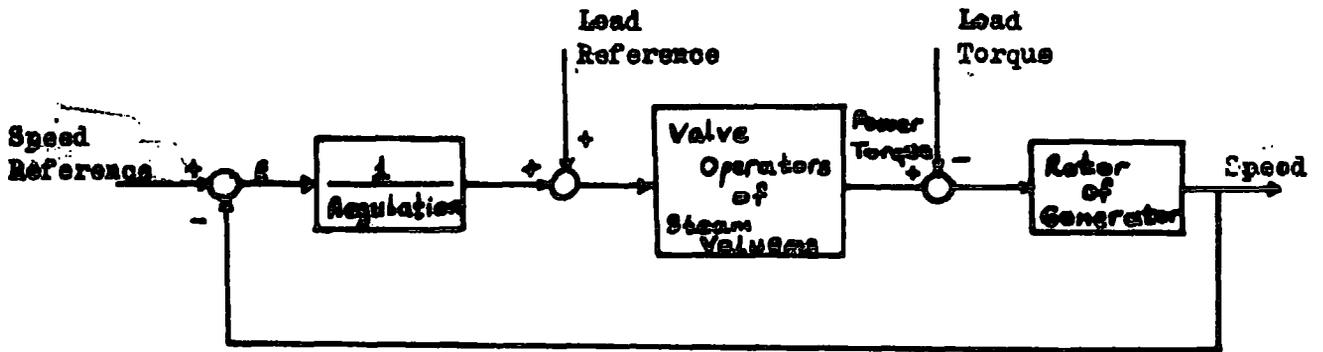


Figure 1.9 Block diagram of "Separate Speed and Load" controls for an electro-hydraulic control system.

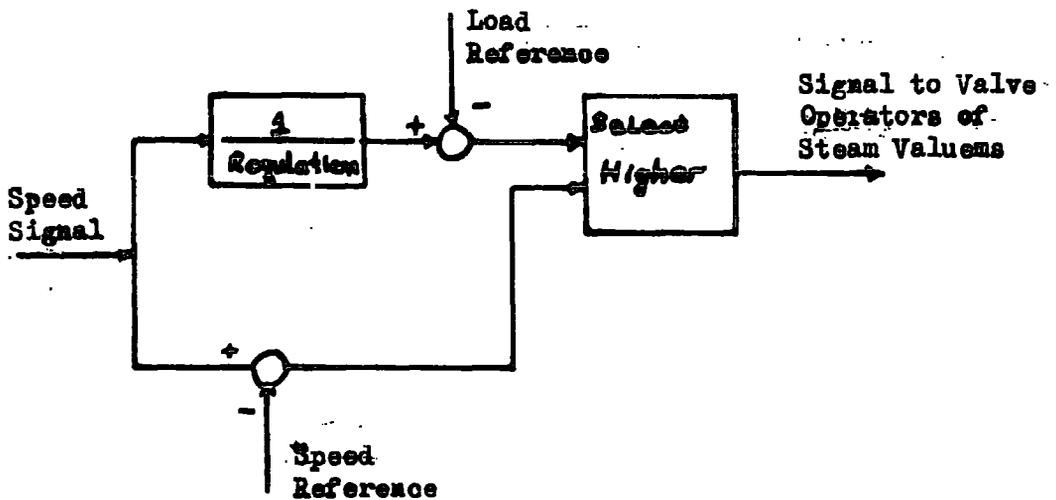


Figure 1.10 Block diagram of "Speed and Load Control by Conditional Loop" system for an electro-hydraulic control system.

CHAPTER II

GENERAL VIEW OF MICROPROCESSOR-BASED CONTROL SYSTEMS

The technical and economic features of microprocessor-based control systems are investigated in this chapter. This investigation has been extended into the use of some technical hardware arrangements. Some of the important hardware configurations for the reliability aspects of on-line computer control systems are also discussed.

2.1 The Convenience of the Installation of Microprocessors in Control Systems

Control equations can be formulated mathematically in a way such that digital computers can perform the same tasks as analog controllers. However, the choice of on-line computer control systems is limited since for single^{loop} control problems they may provide more complex solutions than an analog electrical control system. As the complexity of control problems increases, both analog electrical and mechanical control systems soon become more and more complex whereas this is not necessarily the case for digital controllers. In fact, complexity of the controller directly effects the technical and commercial reasons for the choice of the kind of control system to be used, and both of these reasons are as important in the justification of on-line computer control systems as they are for the justification of the other type controllers.

Microprocessor-based control systems may replace analog

electronic or other forms of hardwired digital electronic, mechanical and computer control systems. [20,3,22,10,12]. Since they are small, fast and have low cost computing power, and they can perform many functions minicomputers do, these systems can improve control and do it for a reduced cost. They offer advantages in performance, flexibility, maintenance, cost and size. [22,12]. They allow the control systems to perform logical operations and give intelligence to control systems, [18], and they are also easy to program. Indirect measurements which are difficult to measure with analog controllers can be obtained in microprocessor-based control systems by calculation, [7]. They can also be used as replacements for special/complex equipment which would otherwise be required to implement the control. They enhance accuracy [22]; especially with digital transducers [13], rather than with nonlinear analog transducers. Digital signals can be easily transmitted without distortion whereas analog signals are prone to error because of their hard-wired couplings and temperature variations. Since microprocessors are very reliable electronic devices and give exact calculations, they provide increased accuracy in control systems. As word length gets longer, accuracy of the machines increases. In the market, 4-bit, 8-bit, 12 bit and 16 bit machines exist; such as 4040 (4-bit), M6800 (8-bit) and CP1600 (16-bit).

Microprocessor-based controllers can be reprogrammed and their hardware reconfigured with minimal physical modification. The same hardware can therefore be used to obtain an entirely different controller for a variety of applications. This sort

of modification increases the system flexibility. Since microprocessors are much simpler than full size computers and can replace a number of analog controllers, as a central element in a control system [25], it is easier to locate a failure and replace it in microprocessor-based process control areas.

Microprocessors are inexpensive electronic devices. They can be used to provide a single control for a number of machines. Because microcomputers replace a number of analog control loops, the total cost of the replaced equipment helps to meet a considerable part of the cost of these computers. Because of their high performance and accuracy, in some systems they reduce system power consumption and thus provide an economic advantage, [26]. They also increase the bulk of product since they are so fast [22,12]. All these make them an economic solution for many applications.

In large control systems microprocessors share the control loops and achieve the control under the supervision of a master computer, [18]. In the control hierarchy microprocessors cannot perform the management functions because of their size. [22,18]. Coordinating entire complex systems, generating up-to-date reports, doing cost accounting, etc., require extensive storage and a single microcomputer may not meet these demands.

2.2 Some Hardware Arrangement Techniques Used In Conjunction With Microprocessors

Some authors have made proposals to keep the number of A/D and D/A converters used with one microprocessor down. ^[2,5,20] This approach results in a drop in the amount of the input and output port

addresses necessary. To utilise only one A/D and only one D/A converter for more than one input and one output of a microcomputer, analog input and analog output multiplexors are proposed. These multiplexors get control commands to select the desired channels, Figure 2.1. As either of these two multiplexors receives a channel select signal, the converter connected to that multiplexor will be switched over to that chosen channel. After the selection of the channel, another command starts A/D or D/A conversion, depending on which is required. Thus one input address to ^{the}A/D and one output address to ^{the}D/A are dedicated. Track-hold circuits are proposed to keep the actuator at its latest position until the next data is available. [21, 5, 20].

In a microprocessor-based position controller of machine tools a register which is essentially an electronic device consisting of flip-flops was utilised at the digital side of each digital to analog converter. The first output data waits in one of these registers until the other output data are housed in the remaining registers. Then these data are sent to ^{the}D/A's. This gives an opportunity to apply all output data almost simultaneously, refer to Figure 2.2. The latch registers duplicate their data towards ^{the}D/A's as they receive individual control signals. [21].

A thyristor cycloconverter converts an alternating voltage of one frequency to an alternating voltage of another frequency by opening and closing the switches within the converter in appropriate manner. A microprocessor is used to calculate these triggering instants and to achieve the control of a three pulse-cycloconverter. The equation that gives the SCR triggering instants

needs sine and arcsine calculations. To increase the speed, a table look-up technique is utilised. This reduces the software overhead time. Again for the speed increment, the author used the Advance Micro Devices (multiplication hardware) for 16 bit multiplication in a short time of $3\mu\text{sec}$, [3]. Table look-up techniques and multiplication hardware have been used also by some other authors to meet the demands of fast acting control systems where the process time is prime concern. Refer to [10] and [16].

2.3 Reliability

Considering technical and economic reasons it can be seen that there are many advantages in employing on-line computer control systems. However, since these digital computer controllers are required to perform a multitude of functions, attention must be paid to reliability. In conventional analog control systems, each functional task is performed by a dedicated hardware element and the failure of that particular element ends with the loss of that function. But, if a computer failure occurs in an on-line computer control system, we lose all control. This loss of control cannot be tolerated in most cases. In order to be able to overcome this problem, a number of different ways have been considered. Complete redundancy is one of these approaches. But this is not economic since twice as much money must be spent to obtain a reliable computerised process control.

The fact that the cost of microprocessor has become so low made it possible to employ more than one microprocessor in one control scheme in order that each microprocessor is to perform

only a small number of functions. To do so, one microprocessor failure causes only the loss of the functions performed by this dedicated microprocessor. Thus, the control still exists since the other functions remained as they were. But this way of solution may not be satisfactory, since one microprocessor is obliged to perform only four or eight functions although it is capable of performing many more functions alone.

More than one computer in parallel was used as an alternative method of achieving reliability. If one of them fails the other one assumes control.

A multiple processor method is used where a switchover mode is acceptable, Figure 2.3. Here are, for example, two processors connected to the same inputs in parallel, and the same outputs can be switched over from one processor to the other. [2] .

For the reliability and survivability aspects an aircraft fly-by-wire system uses four computer and voter units. When a failure occurs on the computer that has been doing the control tasks, the control tasks start to be achieved by one of the other three computers. [6] .

A backup system assumes the tasks to control the system as the computer fails, if manual control is acceptable as the backup mode. After the computer failure has occurred, the operator controls the process manually. In such a control system, only a couple of most important functions may be performed by the operator in order to keep the most important and small part of the system running while the failure is being located and replaced,

Figure 2.4. [2].

Where safety is of main concern all controlling signals should be checked before they have reached the process. In a gunpowder manufacturing plant, two computers are connected in series to perform the control tasks of the hazardous process. One of these computers is a complex one and works at supervisory level. The other one is a simple computer and has the tasks at direct digital control (DDC) level. Mainly the process control system has three distinct control systems; Process Controller or Computer (Supervisory Computer), Digital Control System which may be a microcomputer and Analog Control System. Refer to Figure 2.6. The supervisory computer sends start-up commands to the programmable controller and provides set-point and digital control signal to the analog control system. Since the programmable controller consists of the process shut-down logic and some control logic, it can check the sent commands, for example a command to switch a machine on or off, against its logic which has the status of the process elements. So there is a double check of the commands before any action is taken in the process area. The analog control system, in effect, is a self contained system and can monitor all process variables. In the event of a computer failure, the control of the process is achieved under the operator commands sent from the Manual Interface Panel (semi-automatic control). In fact, a computer failure is indicated by analog values going into alarm and a shutdown of the process computer. [18].

Some vendors introduced another method of backup systems to handle computer failure modes. For this backup system, firstly, manual control is to be acceptable and each output channel must have individual D/A converters. In addition, these output channels can be manipulated by an external source. Under these conditions, the discrete signals inhibit the processor from changing the output. On referring to Figure 2.5, it can be seen that two indicators are used to indicate the input and the output. For a given input the operator can change the output value, raise or lower, manually by using mode switching. In the figure there are four independent channels. Each channel has its own output card that consists of an D/A converter and hold station, so called. [2].

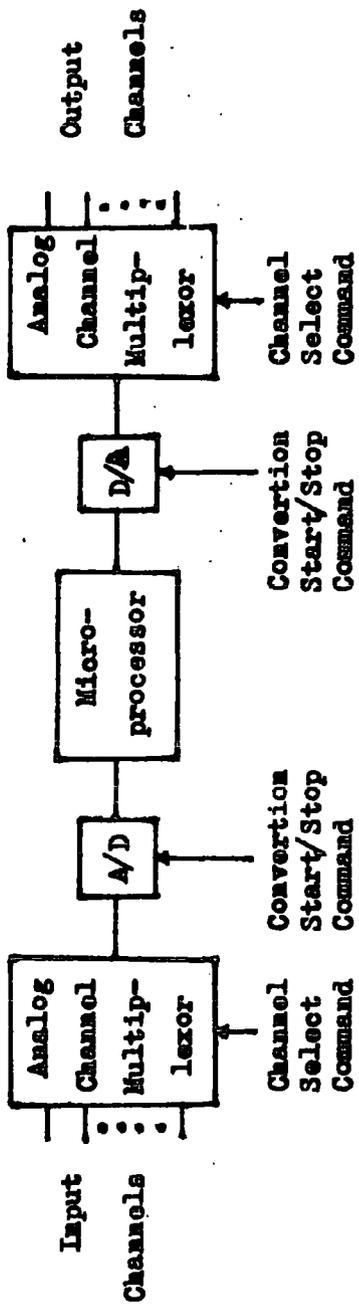


Figure 2.1 'Multiplexing Inputs and Outputs' system

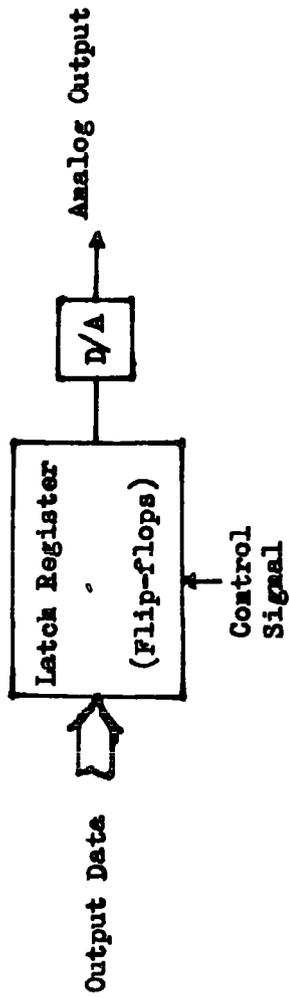


Figure 2.2 'Latching Parallel Data' system

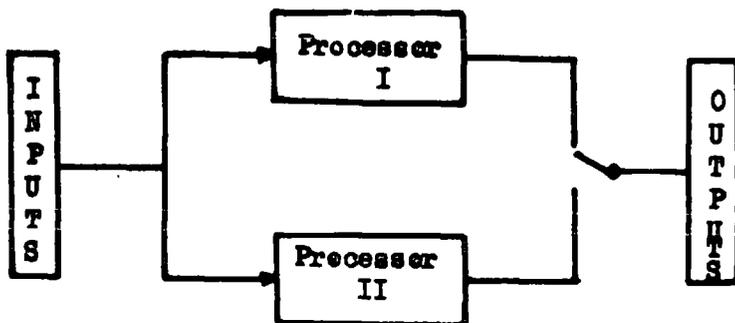


Figure 2.3 'Multiple processors' method

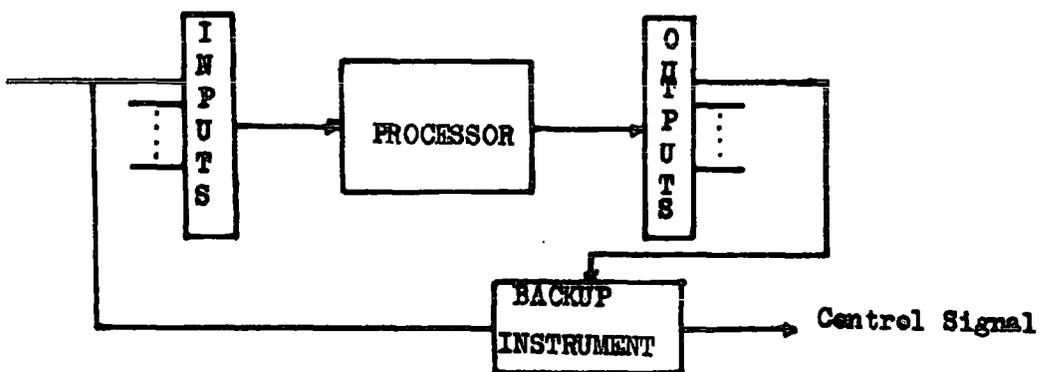


Figure 2.4 Manually backed-up system

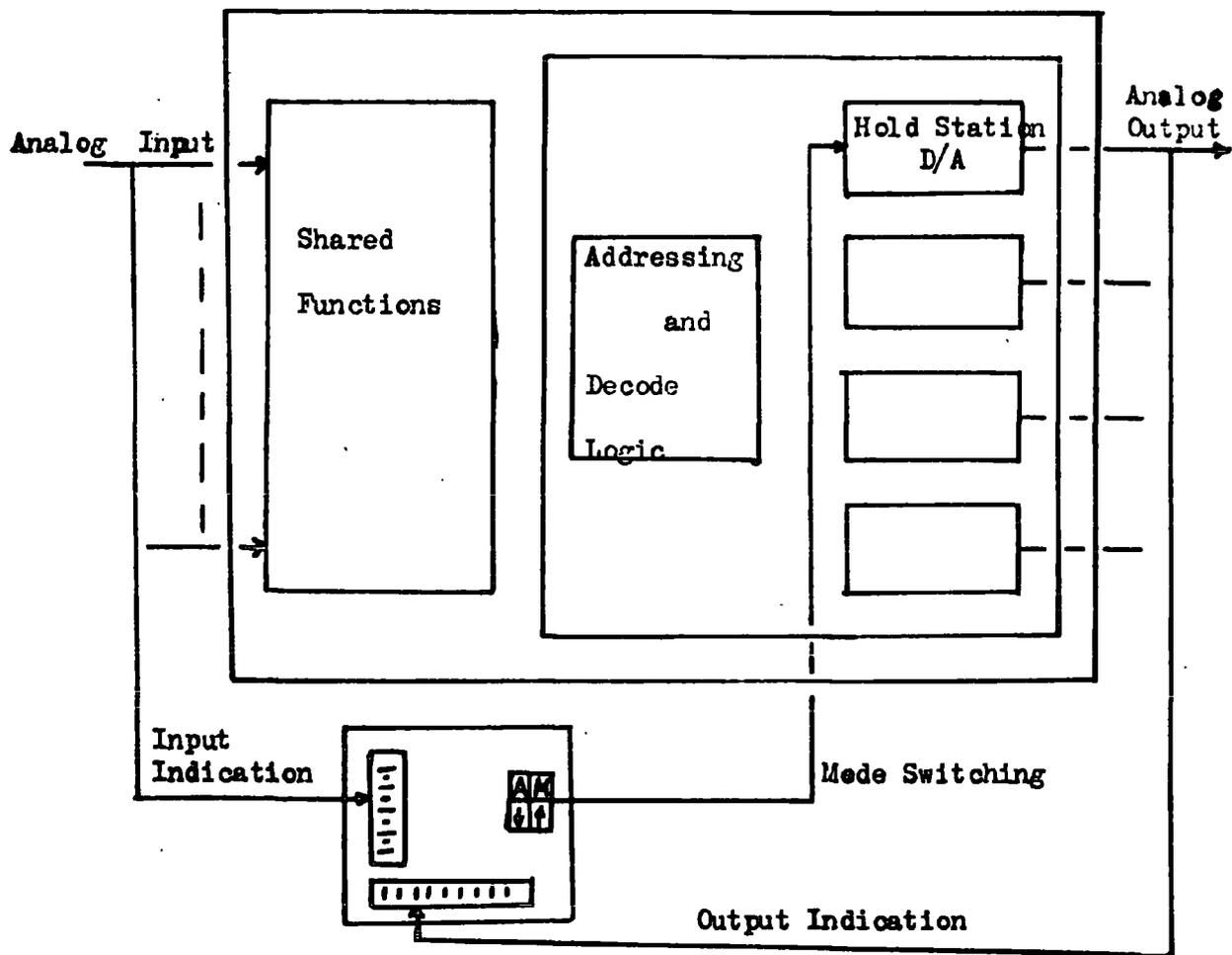


Figure 2.5 Display back-up system

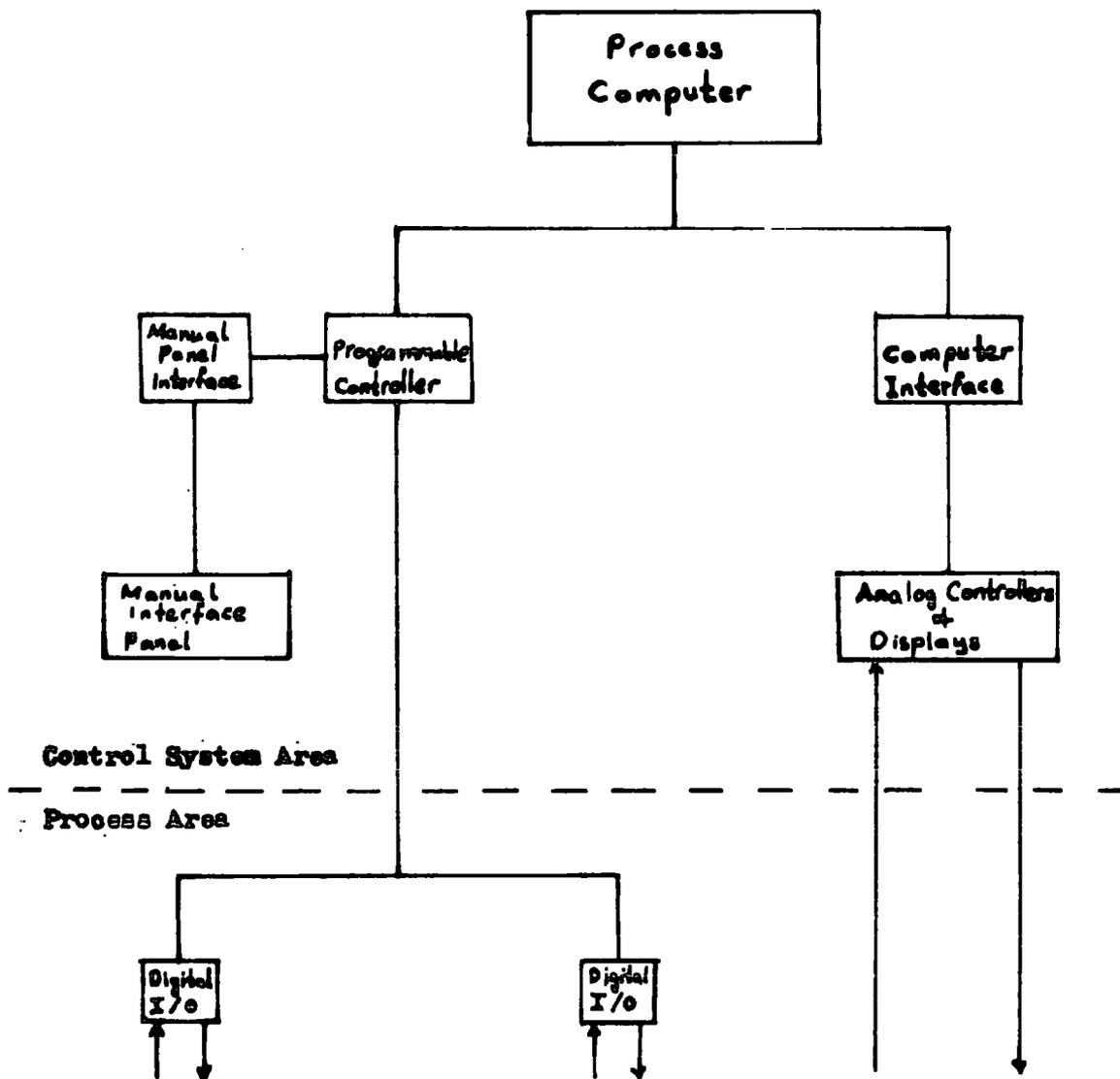


Figure 2.6 An on-line computer control system with two computers connected in series.

CHAPTER III

SWTPC 6800 COMPUTER SYSTEM

3.1 General View of the System

Since this system runs on 16 bit address buses, one MPU can address up to 64 K; in other words memory capacity is 64 K.

Information flow between the system's elements and also between the system and the outside world is done on 8-bit bi-directional data buses. The word length is 8 bit. [24, 14].

It has 72 basic instructions. It is loaded with the machine codes of its assembler instructions. To obtain the machine codes of a program written in assembler code there is a need of a host computer that has the compiler to translate the instructions into machine codes. There are eight addressing modes in the M6800; Dual Addressing, Accumulator Addressing, Inherent Addressing, Immediate Addressing, Relative Addressing, Indexed Addressing, Direct Addressing and Extended Addressing.

A SWTPC 6800 computer system consists of a Microprocessing Unit (MPU), a Read Only Memory (ROM), some Random Access Memory (RAM) s, some Peripheral Interface Adapter (PIA) s and some Asynchronous Communications Interface Adapter (ACIA) s. PIA provides parallel interfacing and ACIA gives serial interfacing. But, in fact, the MIKBUG hardware program in 1 K MC6830 ROM enables a MC6820 PIA package to be used as a serial interfacing adapter. For our project, a MC6820 PIA is used to make serial TTY interfacing to the MPU.

Each of the M6800 family elements in this computer system operates on a single five-volts power supply.

A clock is essential for the processor and the interface circuits. The system may be expanded up to eight interfaces and the various baud rate (one baud is so many bits' flow per second) of each interface circuit can be selected by a bit rate generator such as MC14411.

Figure 3.1 can give some idea of the connections of the system's elements in blocks. Figure 3.1 also consists of a crystal controlled oscillator used as a clock source. The control bus in the figure is shown bi-directional. In fact, some control lines are only inputs to the MPU and some of them are only outputs from the MPU. In addition, all control lines connected to memories have the signal flows from the MPU to the memories only.

3.2 MC6800 Microprocessing Unit (MPU)

This is a Large Scale Integration (LSI) device and consists of two 8-bit accumulators, one 16-bit Index Register (X), one 16-bit Program Counter Register (PC), one 16-bit Stack Pointer Register (SP) and one 8-bit Condition Code Register (CC). The number of the program instructions used to activate the logic and arithmetic functions of the MPU is 72. It contains instruction decoding logic, Arithmetic and Logic Unit (A.L.U.), and program sequence control. It is provided with an 8-bit bi-directional data bus, a 16-bit address bus and some control lines; Read/Write (R/W), Valid Memory Address (VMA), Data Bus Enable (DBE), Interrupt Request (IR_Q), Restart (RES), Non-Maskable Interrupt (NMI), Go/Halt (G/H), Bus Available (BA) and Three State Control (TSC). It is also provided with a two-phase

clock whose operating rate is up to 1 MHz. The MPU uses the clock as a timing reference to execute instructions. For example, the MPU places an address on the address bus during one phase of the clock and the data bus will be active during the other phase of the clock. Refer to Figure 3.2.

The MPU transfers information between the memory units and the outside world so that if it is required the MPU will fetch an item of data from a memory address and then store it to the desired output. The sequential fetching of instructions in the program memories is done through the PC. After loading a program and giving the starting address to the PC register, the Go (G) command is applied to start the execution of the program. The processor then loads the address in the PC on to the address bus and the Read pulse (high state of the R/W signal) strobes the data at the given address into the MPU. The instruction decoding logic of the MPU will enable the MPU to interpret the strobed data as an instruction or just a number. The execution of instructions is done in the A.L.U. After fetching a data the value of the PC will automatically increase by one. The new value in the PC is the next address for the MPU to get the next data to execute. However, if the program requires the result of an execution to be stored in a memory location, then the MPU places the address, where the resulting data is to be stored, on the address bus. The output signal of an AND whose inputs are the VMA and ϕ_2 clock signals will inform the external devices of the MPU that there is a valid address on the address bus. In fact, this enable signal is applied to the

Enable lines (E) of the memories and I/O devices. The DBE which is normally the $\phi 2$ clock signal puts the data on the data bus from the MPU.

The IRQ line of the MPU gets signal from PIA/ACIA. An interrupt signal will be sent to this line if there is an available data on a peripheral while a program is running. Next, the present contents of the PC, X, accumulators and CC register will be stored in seven bytes of RAM starting with the memory address that the Stack Pointer (SP) contains, and proceeding in descending order of memory addresses downwards in sequential manner, if the interrupt mask bit in the processor condition code register is not set. Then the MPU sets the interrupt mask bit to ensure that it will not response to any further interrupt before the completion of the serving to the present one. The MPU starts running a program that serves the IRQ. The starting address of the subroutine is loaded into the PC from two memory locations. The instruction RII (Return from Interrupt) causes the MPU status loaded with their contents before the interrupt occurred. Upon completion of the service program the execution of the suspended program is resumed. In fact, this way of data transfer from the outside world may save processing time if it is compared with the way of data transferring where a program would periodically examine every single port in turn for an available data.

The RES signal recovers the MPU from a power failure and it may also be used for an initial start-up of the processor. To be able to communicate with the MPU through a teletype while there is a program continually this signal is to be applied. That results

in stopping the program being executed. The RES signal is also connected to the RES line of PIAs. This signal has the effect of setting all PIAs' registers to logical zero.

The Go signal (high state of the G/H signal) starts the execution of program at the address in the PC. If this signal is at low state program execution will be halted.

In the high state the BA signal indicates that the MPU has stopped execution and the address bus is available.

The NMI signal has no effect of the interrupt mask bit in the CC register. At the presence of this signal the MPU status are stored away through the SF. The interrupt mask bit is set. The MPU branches to a routine that serves this interrupt. The address of the routine is stored in two memory locations and the CC is loaded with this address automatically. Upon completion of the routine, the previous program is resumed by the MPU.

3.3 MC6830 Read Only Memory (ROM)

This LSI device has static operation. It is a 1024 x Byte = 1K byte package. It has the MIKBUG hardware program in it and MIKBUG is unalterable. The MIKBUG monitor program does not only provide the user with the subroutines to use in his programs but also enables him to examine a memory location and change it at will (except the memory locations of ROMs), and to start execution of his program, and to load an object tape, and to print a block of memory via a keyboard.

MIKBUG also uses one Random Access Memory package as a temporary data storage.

3.4 MC6810 Random Access Memory (RAM)

This is used to store software programs. It is an alterable Read/Write memory. Each RAM is organised as 128 Bytes.

3.5 MC6820 Peripheral Interface Adapter (PIA)

The MC6820 Peripheral Interface Adapter is a parallel type interface adapter circuit. Its MPU side has 18 lines provided. (Refer to Figure 3.3). Eight of these lines form the bi-directional data bus, and the others are used for addressing and controlling the PIA and the internal registers of the PIA. The peripheral side has 20 lines and 16 of them form two 8 bit bi-directional data buses. Four out of these 20 lines are utilised as control lines.

The address of PIA is defined by hardware selection logic. The Chip Select (CS0, CS1, CS2) lines on the MPU side are used to select a PIA. The PIA is programmable. Each PIA has six 8-bit registers in two sets called A side registers and B side registers. Each side has one Data Direction Register (DDR), one Output Register and one Control Register. Which set will be utilised as input or output depends on the bit positions of the DDR.

The MPU treats the DDRs and the Output Registers on each side as a single memory location and the MPU treats one PIA as four memory locations.

In conjunction with the Register Select (RS0 and RS1) lines one bit of the control register directs the MPU to the DDR or Output Register.

On the MPU side of the PIA there are two IRQ lines; one of them is used for the interrupt request of one side and the other one for the interrupt request of the other side.

Two control lines CA1 and CB1 on the peripheral side are only input. The CA2 and CB2 may, however, be programmed to act as the peripheral outputs or the interrupt inputs.

Since the PIA is a parallel I/O device, to connect an analog signal to a PIA an Analog to Digital (A/D) converter circuit is to be placed between the signal and the PIA's data lines. The digital outputs of an 8 bit A/D converter are connected with these 8 bit data lines. Since a convenient bit pattern of the CRA (CRB) may set the interrupt flag of the same control register during a high to low transition on the CA1 (CB1), the CA1 (CB1) control line of the PIA may be connected to the status signal (the signal that informs that the A/D conversion has been completed) of the A/D converter circuit. Since setting one of the control register bits may make CA2 (CB2) go high, the CA2 (CB2) may be connected to the "Start Conversion" line of the A/D in order to make the A/D circuit start the conversion.

Firstly a software program loads the control register with a bit position set so that the CA2 (CB2) goes high and consequently the conversion starts. Upon completion of the conversion the status signal of the A/D goes low, so does the CA1 (CB1) and the program may read the data into the MPU after the recognition of the IRQ flag that was set by high to low transition on the CA1 (CB1).

3.6 MC6850 Asynchronous Communications Interface Adapter (ACIA)

This M6800 family element enables the user to make serial data communications with the MPU. The user's TTY control terminal, Keyboard and Cassette Recorder may be interfaced to the MPU through ACIAs. The Motorola's MIKBUG hardware program in 1 K

MC6830 ROM enables PIA to be utilised for the same purpose.

Since we are using a PIA in the project to make serial communications, the references [24], [14] and [15] should be referred to obtain the detailed information about ACIA.

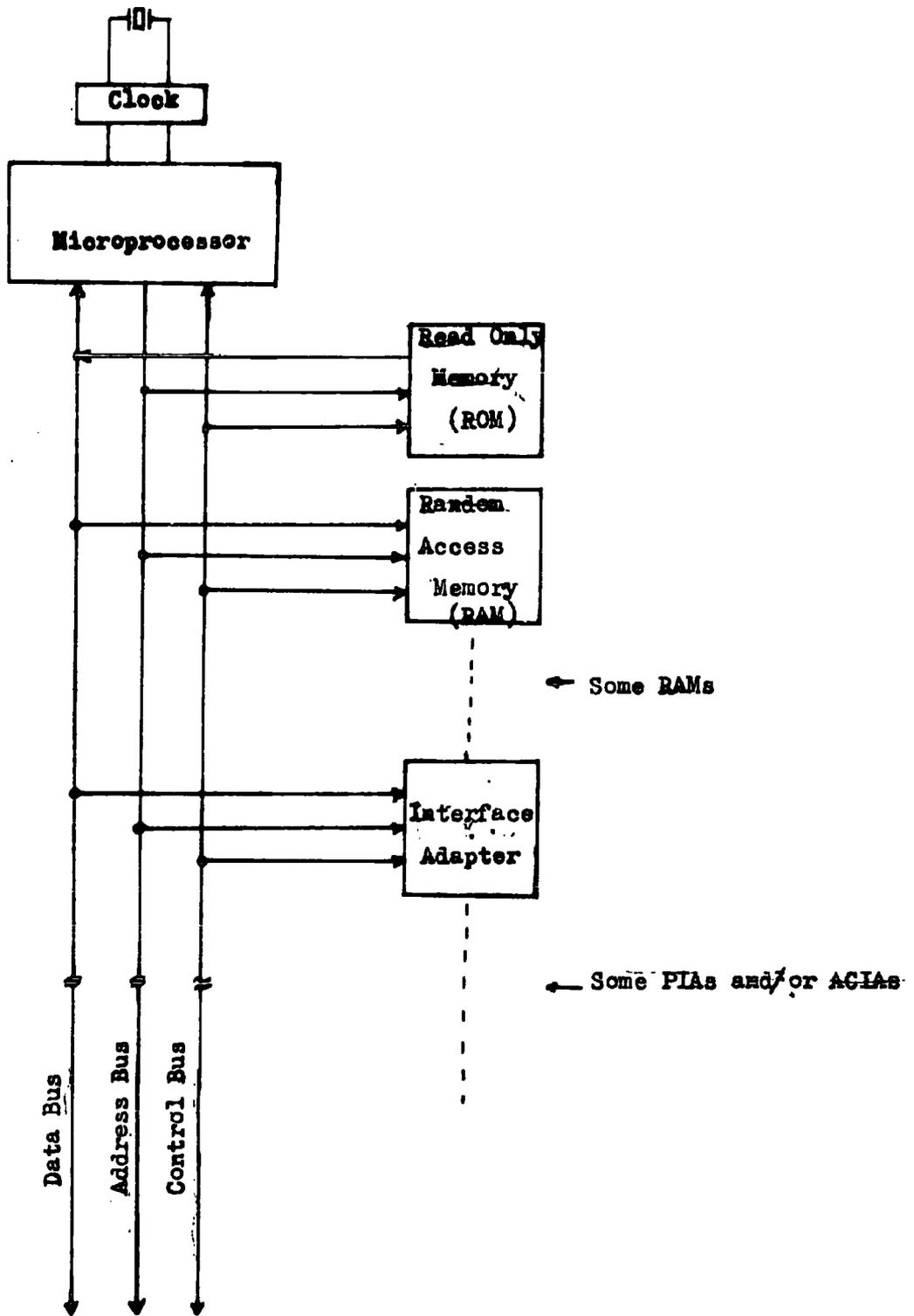
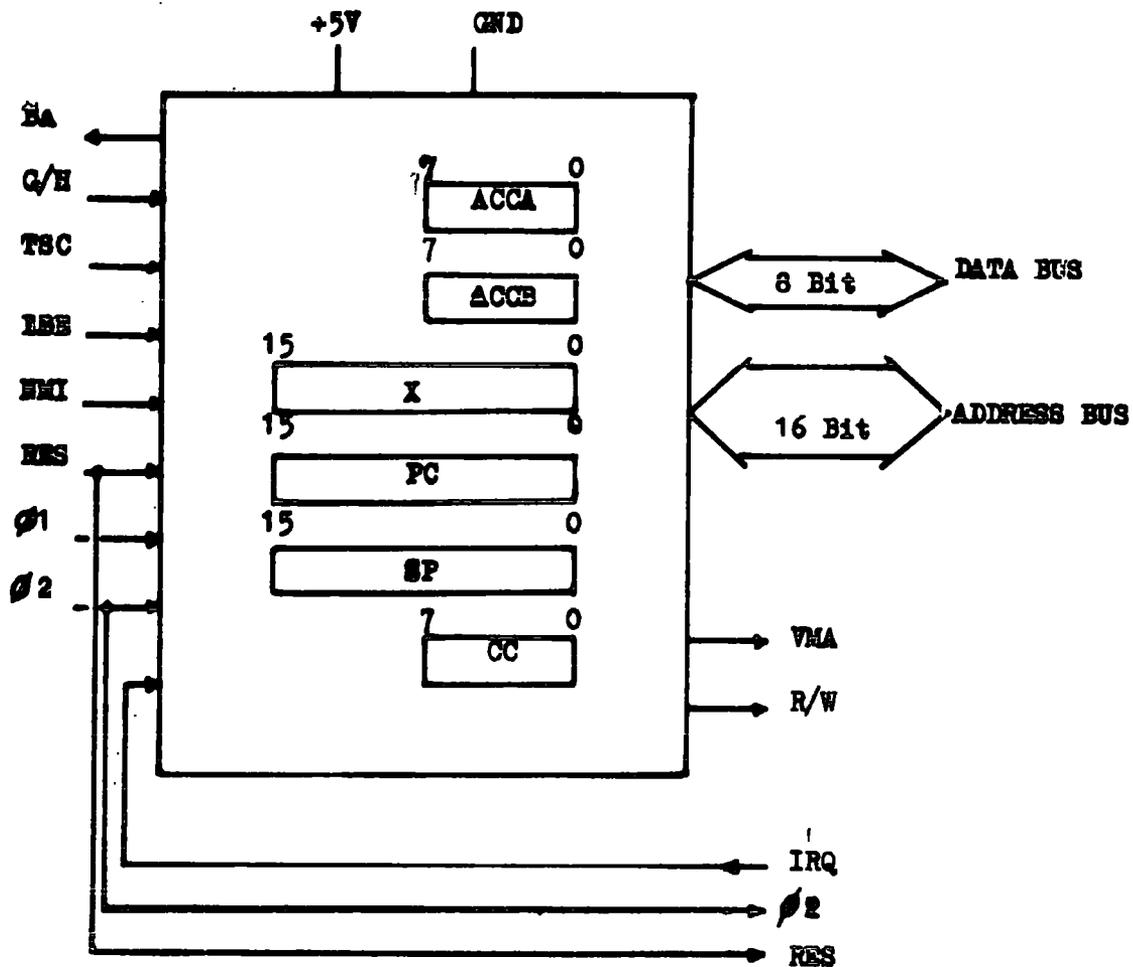


Figure 3.1 M6800 Microcomputer family block diagram



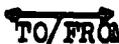
 TO/FROM	 TO/FROM
6800 CONTROL CIRCUITS	MEMORY AND PERIPHERALS

Figure 3.2 MC6800 Microprocessing Unit

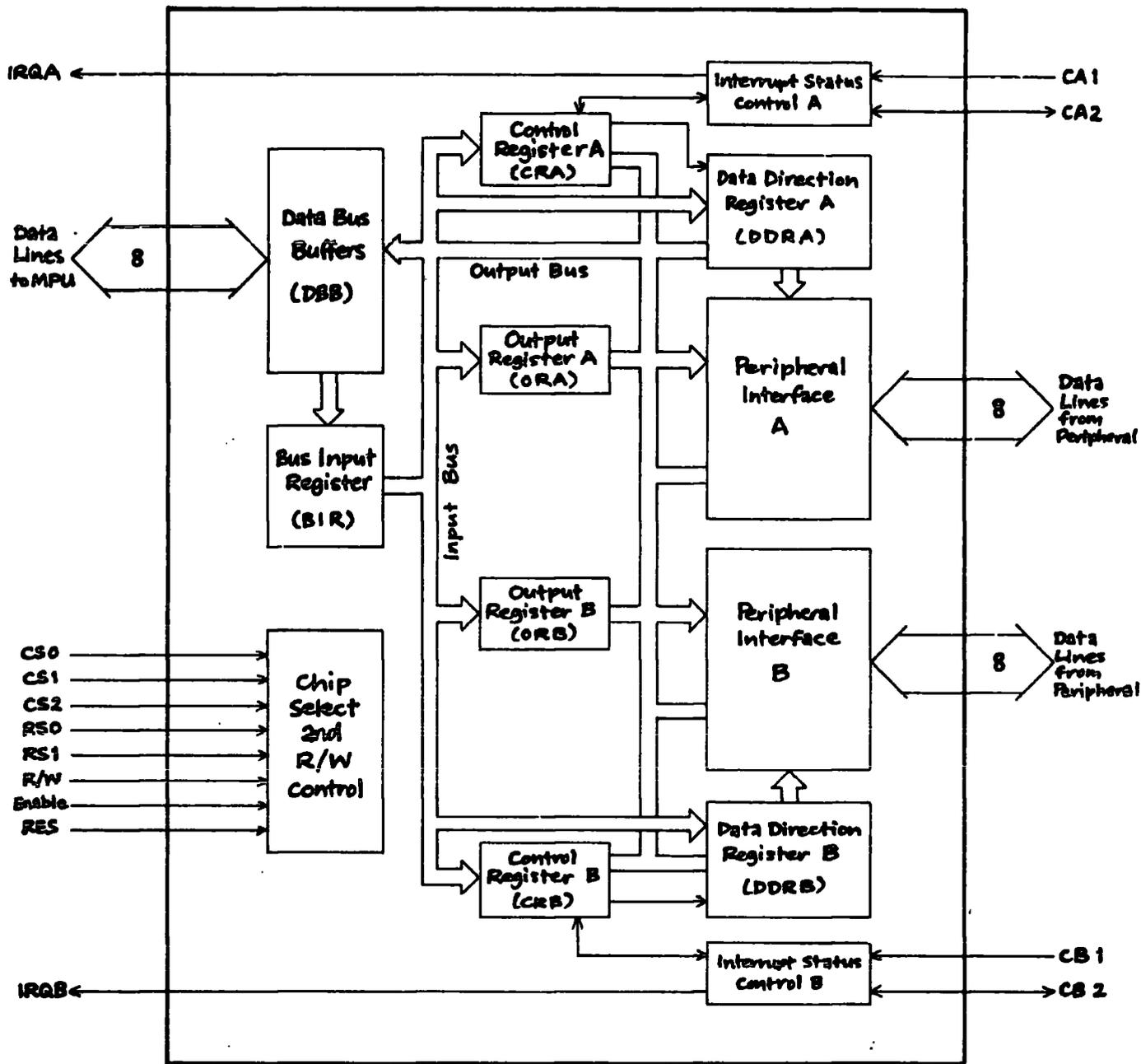


Figure 3.3 The expanded block diagram of MC 6820 Peripheral Interface Adapter

CHAPTER IV

EXISTING GOVERNOR SYSTEM AND ITS

ANALOG COMPUTER MODELLING

A valve controlling system block diagram supplied with the step-by-step on-load valve testing procedure was obtained from a manufacturer. This chapter mentions the brief description of this electro-hydraulic valve control system operation and the valve testing steps. The reduced governor model with some changes in the block gains and time constants, and the mathematical model of this system established on the analog computer are also given in this chapter. The block diagram calculations for the analog computer simulation, however, are in Appendix A.

4.1 The Governor Specifications

The block diagram of the given Electro-Hydraulic Governor system is in Figure 4.1, and the gains and the time constants given by the manufacturer are listed in Table 1.

The Power Piston response in the opening direction and the Actuator response in the closing direction for a 6 volt step input also given by the manufacturer is illustrated in Figure 4.2.

This Governor achieves the control task for both Control and Intercept valves.

GAINS		UNITY	TIME CONSTANTS in seconds	
G1	5.38	milli-amps/volt	T1	.04
G2	2.666	<u>cubic inches flow</u> milli-amp	T2	.013
G3	1.25	1/Area squ.inches	T3	.16
G4	20.	10Volt/Stroke inches	T4	.0028
G5	1.		T5	.1
G6	.0755	1/Area squ.inches	T6	.1
G7	1.666	10Volt/Stroke inches	T7	.1
			T8	1.

Table 1.

GAINS		UNITY	TIME CONSTANTS in seconds	
G2	2.66	<u>cubic inches flow</u> milli-amp	T5	.1
G3	1.25	1/area squ.inches	T6	.1
G4	2.	10Volt/Stroke inches	T7	0.0
G5	1.		T8	.01
G6	.0755	1/Area squ.inches		
G7	55.	10Volt/Stroke inches		

Table 2.

4.2 Brief Description of The System Operation

The Phase Rectifier with the gain G_4 is a Linear Variable Differential Transformer and the position transducer with the gain G_7 is a rotary type transducer. Since the Primary Actuator and the Power Piston are pure integration devices, to obtain a constant output position of these devices for a constant input signal and to enhance their performance, the feedback paths are established around these devices.

The Servo-Valve Amplifier is an electronic device and amplifies the error signal. The current supplied by the amplifier operates the Servo-Valve. The oil flow from the Servo-Valve to the Primary Actuator results in the piston movement of the Actuator. Since the Pilot Valve piston is mechanically in connection with the Primary Actuator piston, the Pilot Valve operates and a high pressure oil flows into the Power Piston. The valve operation is achieved over a lever that links the valve to the Power Piston.

The gains of the Servo-Valve, Primary Actuator, the rectifier gain G_4 , Pilot Valve and Power Piston may vary between contracts where different pressure requirements exist. The gains G_1 , G_7 and also G_6 in Figure 4.1 vary between Governor and Intercept valves.

4.3 The Actual On-Load Valve Test Sequencing

4.3.1 Purpose of the Test

The on-load valve testing is achieved by doing the following in sequence: selecting the valve to be tested, adding a negative going ramp signal to the operating Governor input signal and consequently closing the valve, tripping the valve, applying a positive going ramp signal to the Governor input, re-application of the negative going ramp, releasing trip and bringing the valve back to its pre-test position.

During the test a permanent record of the valve performance is provided by an X-Y plotter. The X input of the plotter receives the valve position output and the Y input is in connection with the Governor input. The graph plotted during the test is the input versus output curve. The curve characterises the behaviour of the Governor by indicating if the trip operation has failed and if the valve being tested gives unexpected response because of the erosion on the valve. Getting no vertical trace is the result of the fact that there is failure of the trip operation, because the plotter's X input is expected to be zero and the Y movement still exists during the ramp operation following the tripping of the valve. Any friction or faulty response of the valve will cause^a a kink in the traces being obtained during the valve closure and re-opening.

The newly recorded response for each valve is checked against its previous test record and the differences are observed to gain ideas on the present conditions of the trip operation system and the valve itself. Then the necessary precautions may be taken to avoid risk.

4.3.2 Basic Specifications of the Used Pushbuttons

The on-load tests on the valves are carried out remotely from a control room.

There are a certain number of pushbuttons on the control panel dedicated to the same number of the valves on the turbine. These buttons are called "Select Pushbuttons". One valve is tested and one plotter record for this valve is obtained at a time. When pressed the button initiates the test procedure on the appropriate valve. The Select buttons are latch-down type of buttons.

Also there are two non-latching pushbuttons that have to be pressed and held down in sequence during the tests. These buttons are common to all valves. One of these non-latching ones called "Test"

initiates a command to interrupt the present Governor signal when pressed and causes the restoration of the Governor signal when released by means of the ramp signal. The other button called "Trip" is utilised to trip the valve under control.

The Select and Trip buttons also operate a set of switches so that only one valve can be selected at a time and the Trip button cannot be depressed before the Test button.

4.3.3 Detailed Test Procedure

The on-load valve test is done step by step as follows:

Select Valve

The valve subject to the test is selected by pressing the button dedicated to that particular valve. The signal obtained by pressing the button results in switching the X input of a plotter from its normal valve position input to another X input which in turn gives the X movement during ^{the} encircling ^{of} the valve number on the plotter. The Y input of the plotter is also supplied by the same module after the SELECT command has ^{been} reached. After the valve identification, the relay is released and the X input on this relay is switched to the valve position input. Another relay called "Fault Reset Relay" is also energised and this provides a signal to the Governor control panel to indicate "on-load test" in progress.

Press Test Pushbutton

When this test button is pressed down a ramp signal is added to the present Governor input signal and the valve closes.

During the valve closure, the system automatically checks the signal selection (voting) circuit in the Servo-Valve Amplifier.

Press Trip Test Pushbutton

The Trip Test Relay is energised, this results in tripping

the valve. To avoid accidental rapid valve closure that would result by depressing the TRIP button before the TEST button, here some switches are used as interlocks.

Release Test Pushbutton

The close Governor signal is cancelled. The Primary Actuator operates but no valve movement occurs. A limit switch closes to guard against accidental rapid operation of the valve.

Press Test Pushbutton

The close Governor signal is applied and the Actuator closes down. The valve still stays at its close position.

Release Trip Test Pushbutton

The Trip signal is cancelled.

Release Test Pushbutton

The close Governor signal is interrupted and the valve re-opens.

Release Selector Pushbutton

The fault reset relay is de-energised and the "on-load test" lamp on the Governor control panel is extinguished.

4.3 Simulation of the Actual Governor

For simplicity, the given Governor model was reduced to the system whose block diagram is illustrated in Figure 4.3. The gains and time constants of the blocks used in this reduced system have been listed in Table 2. The block diagram calculation of the system for the analog computer simulation is given in Appendix A.

When the calculated blocks were connected in their order the mathematical model for the analog computer became as illustrated in Figure 4.4.

For better observation of the system response the amplitude of the step input was taken to be as big as -26.25 volts. The negative signal is used to obtain the positive outputs of the amplifiers 2A2 and B4. The Power Piston response in^{the} opening direction and the Actuator response in^{the} closing direction for -26.25 volts have been plotted and seen to be as in Figure 4.5. As it can be seen in the figure, the calibration has been done and the Position Output has been multiplied by ten. To compare the actual system responses of the Actuator and the Power Piston in Figure 4.2 with their model responses, a further scaling has been done by switching the capacitor value of the amplifier B4 from 0.1 μ F to 0.01 μ F and feeding the output signal of the pot P7 to the summing amplifier 2C5 over 1M Ω but not 0.1 M Ω . The Actuator and the Power Piston responses obtained after the scaling are illustrated in Figure 4.6. The actual system responses can be easily compared with the model responses in Figure 4.7. This figure is made up of Figure 4.2 and Figure 4.6.

The model response is ten times slower than that of the actual Governor.

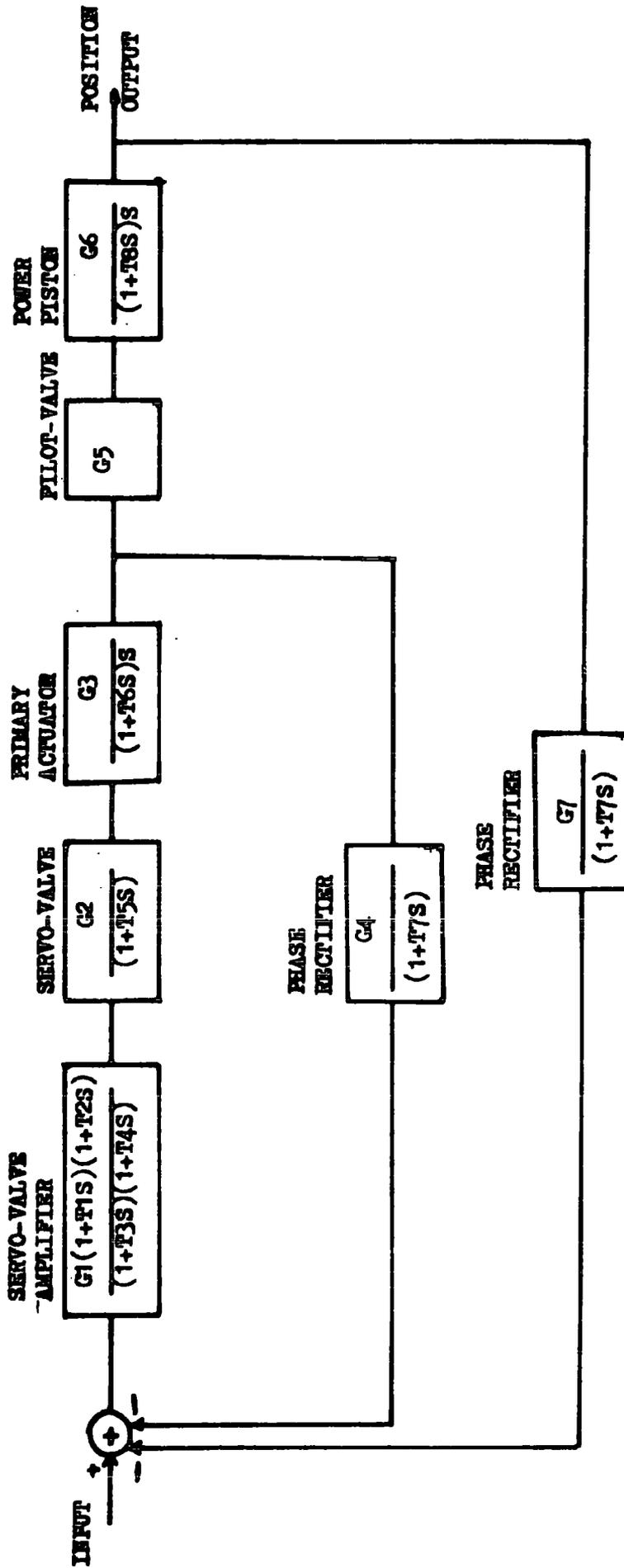


Figure 4.1. Basic block diagram of the present Governor.

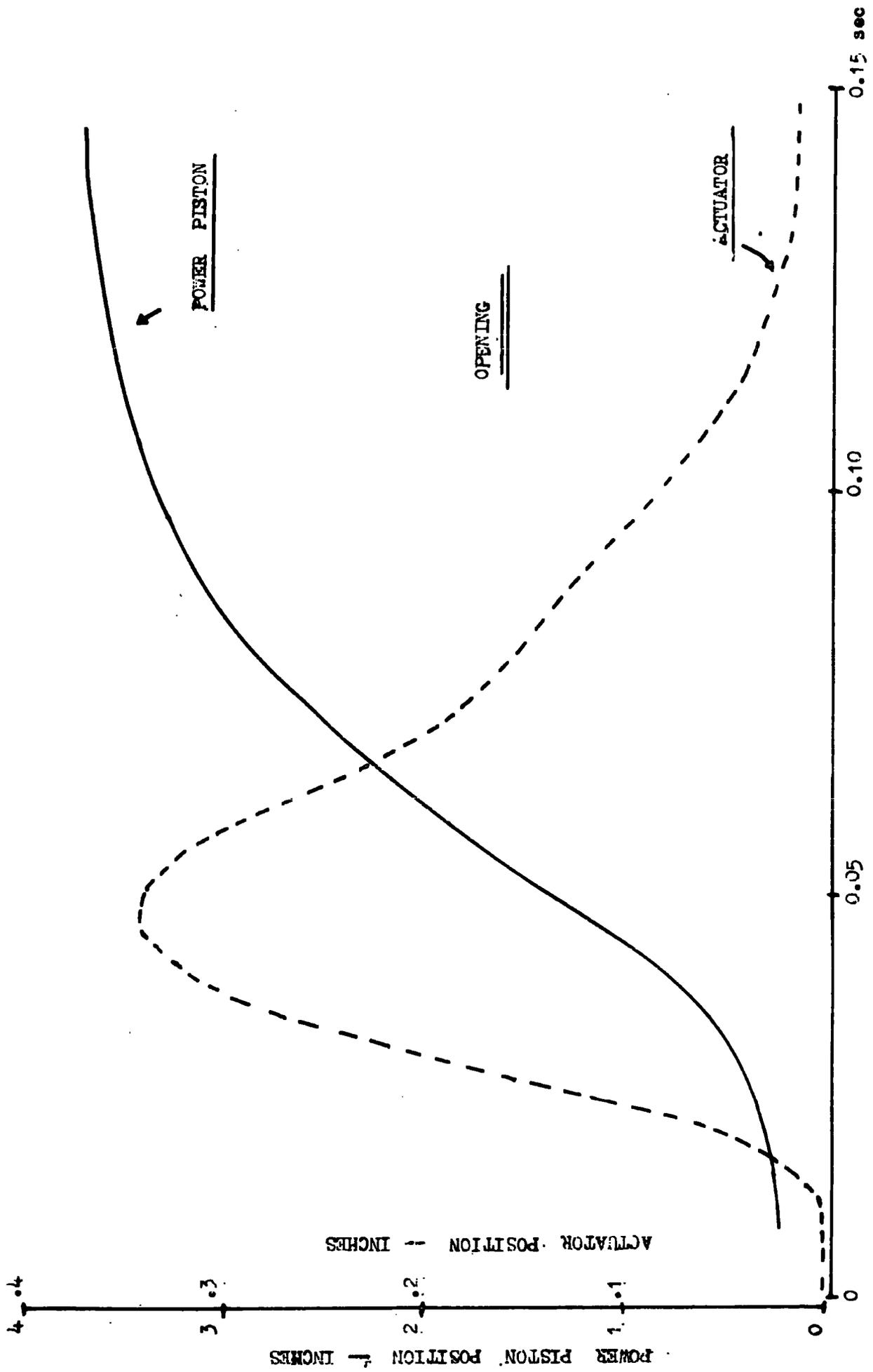


Figure 4.2 Power Piston response in opening direction and Actuator response in closing direction to 6 Volt step input.

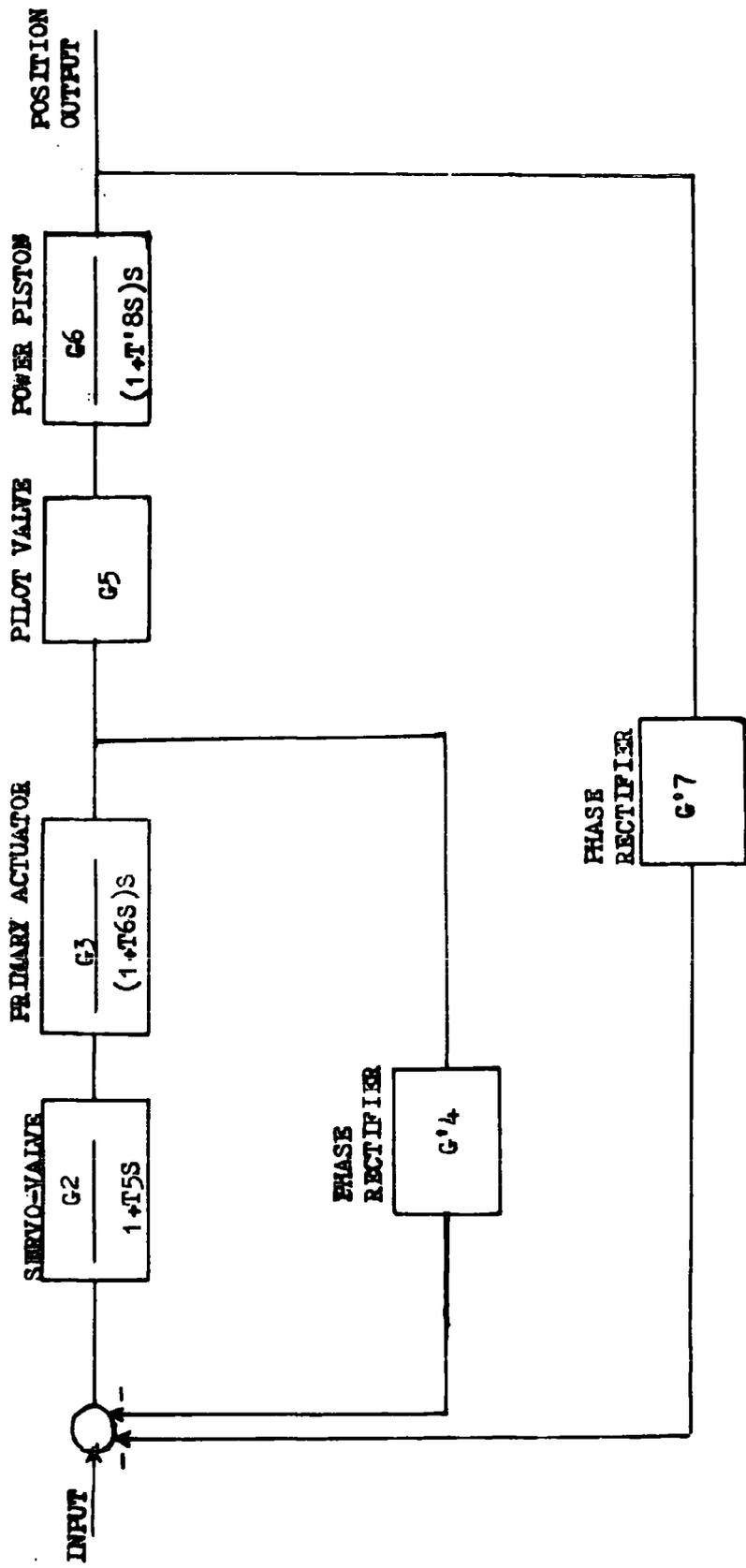


Figure 4.3 Reduced governor model

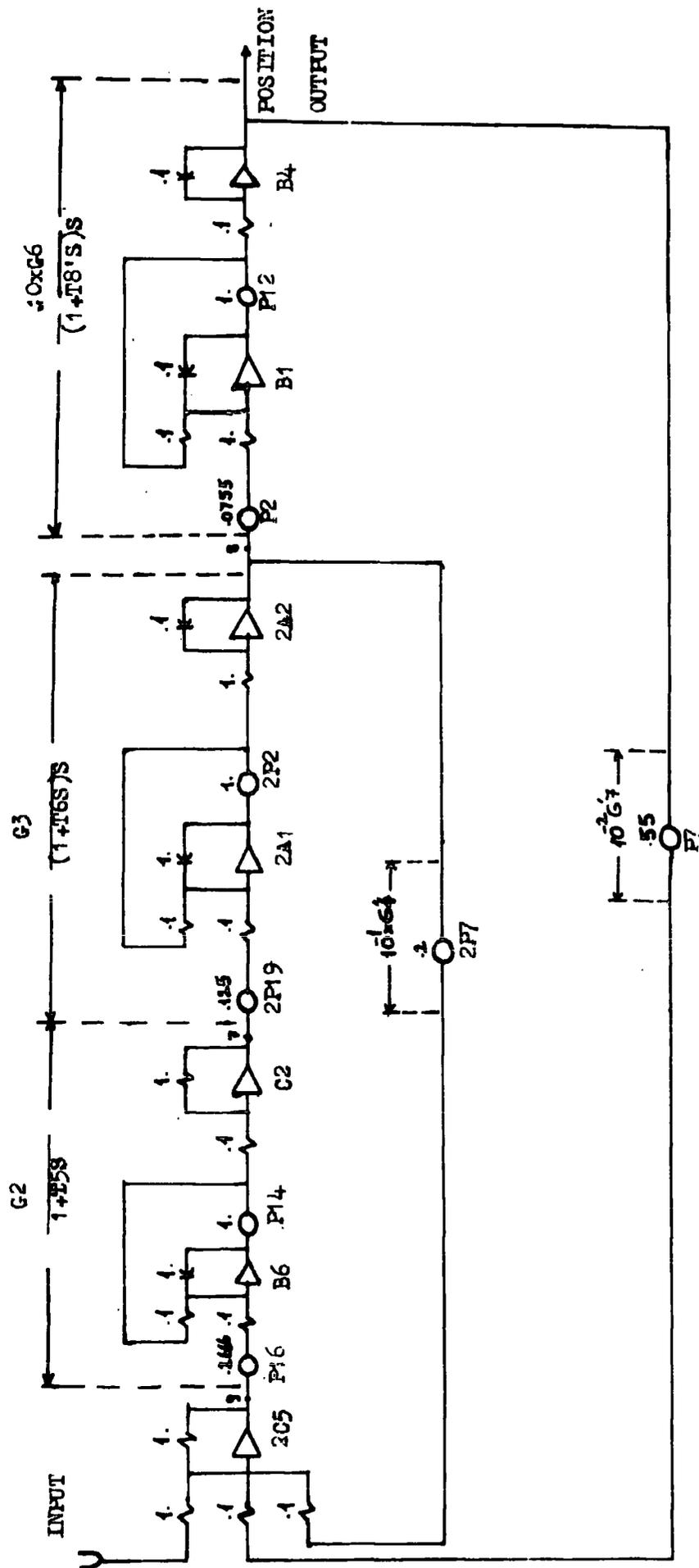


Figure 4.4 The mathematical model established on the analog computer

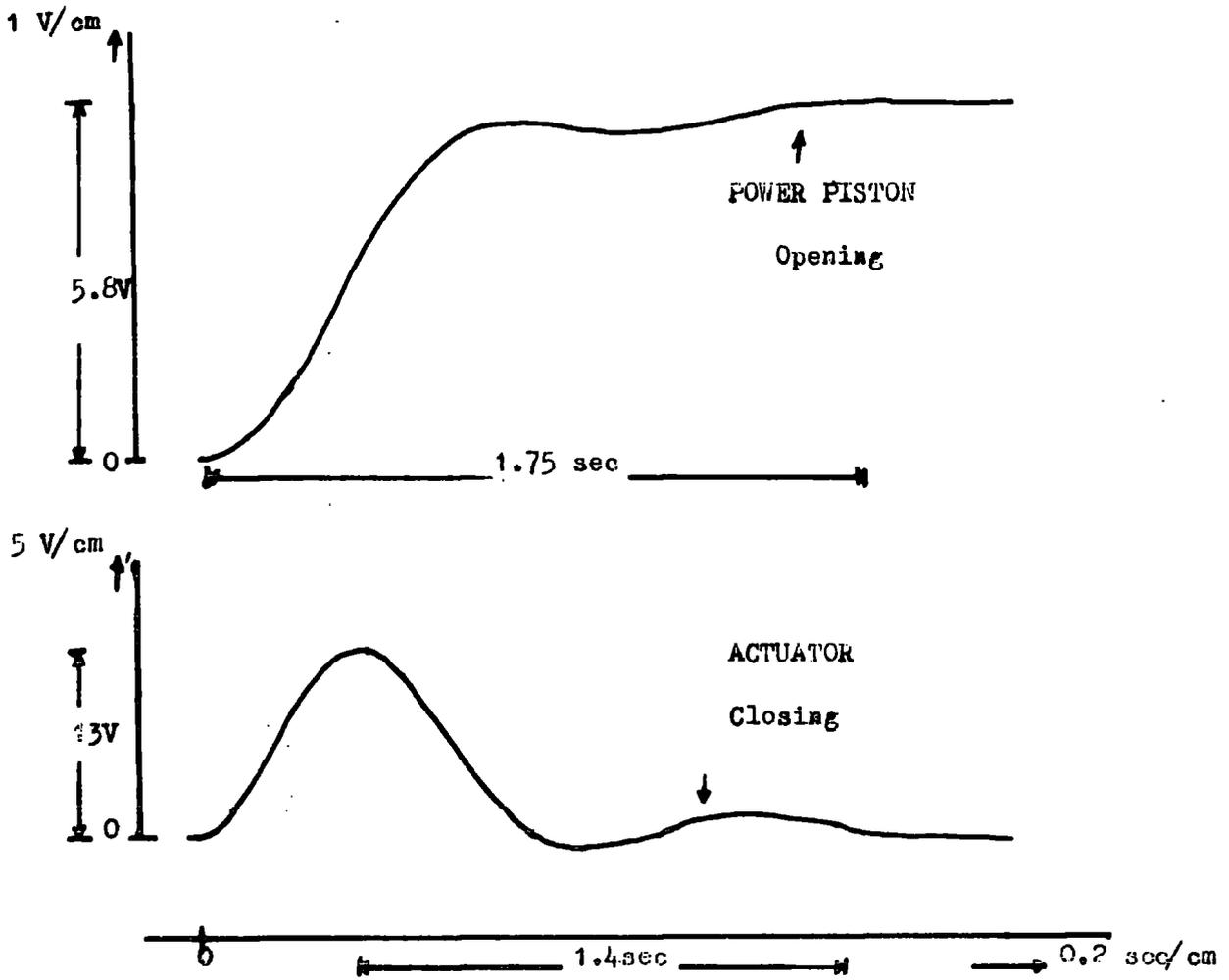


Figure 4.5 Power Piston response in opening direction and Actuator response in closing direction to -26.25 volt step input

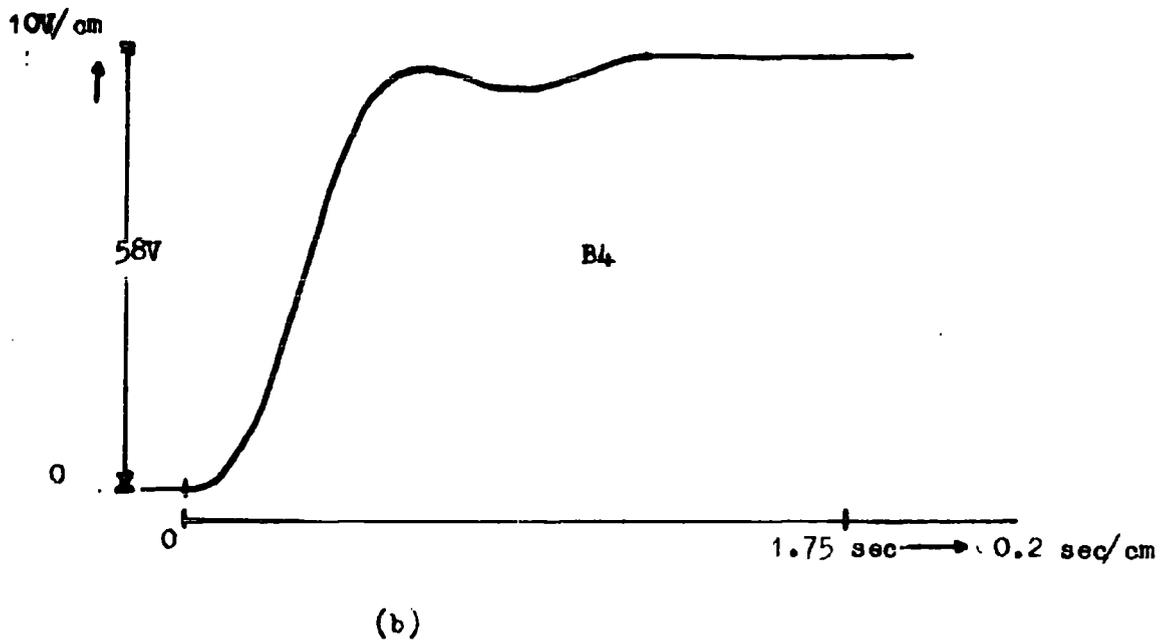
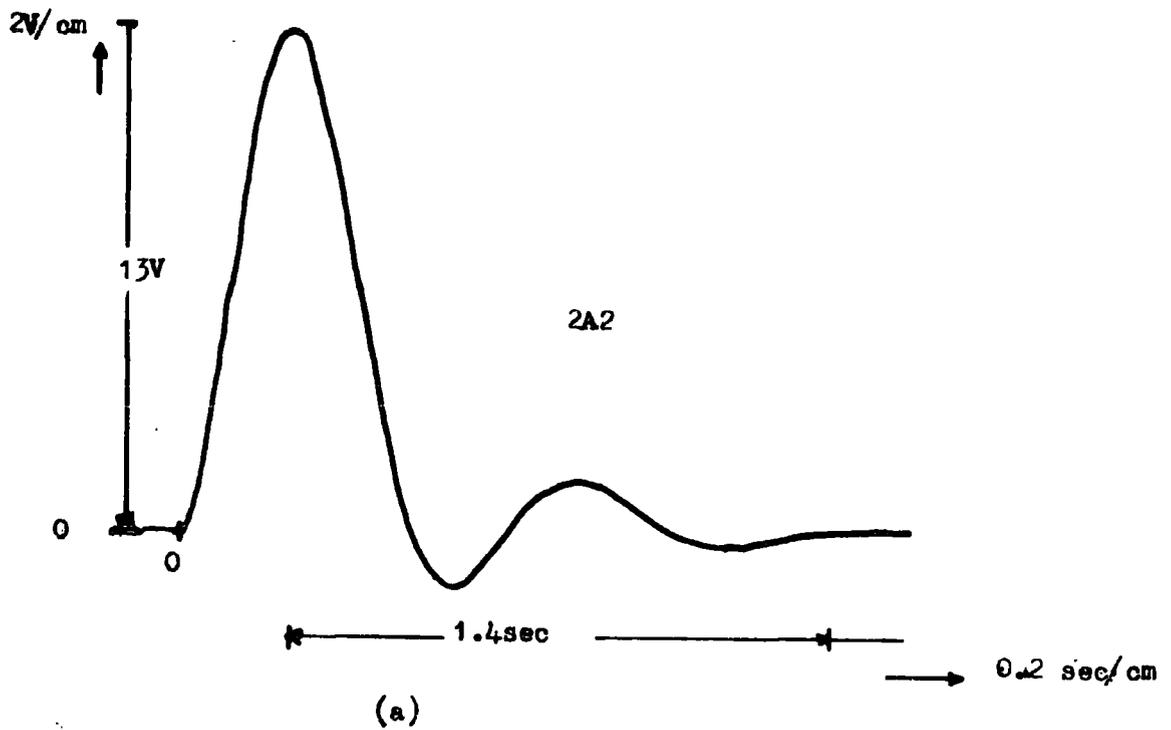


Figure 4.6 Response of the mathematical governor model to -26.25° step input
 (a) Scaled response of the Actuator model in closing direction.
 (b) Scaled response of the Power Piston model in opening direction.

Power Piston Position - Volts
 Actuator Position - Inches

69.5V 16V 4 0.4
 58V 15V
 52V 12V 3 0.3
 34.6V 8V 2 0.2
 17.3V 4V 1 0.1
 Model Actual

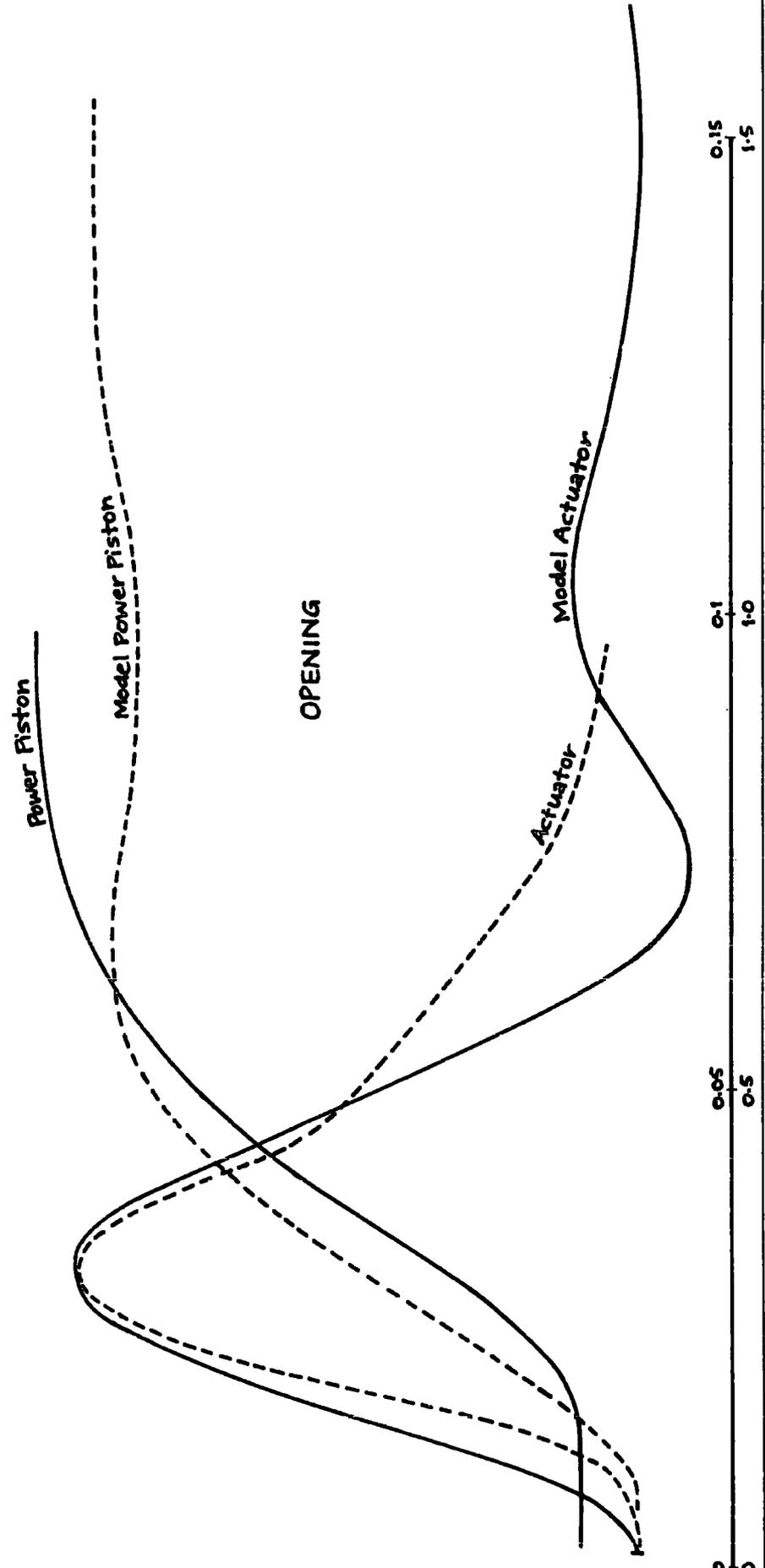


Figure 4.7 Comparable responses of the actual system and its reduced mathematical model

CHAPTER V

MICROPROCESSOR-BASED ON-LOAD VALVE SEQUENCING

Microprocessor-based On-Load Valve Sequencing is developed in this chapter. The chapter describes both the manner in which the MC6800 microprocessor is interfaced to the external systems and the manner in which the microprocessor program is developed. The arrangement of the plotter paper used to obtain a permanent record of valve performance and the time calculation of the ramp signal created via the system software are given in this chapter. [24, 14, 15].

5.1 System Hardware Organisation

The experimental system contains the MC6800 microprocessor, 4096 words of the 128 x 8 RAM (McM6810)s, a 1024 x 8 ROM (MC6830) and three MC6820 PIAs.

The present peripherals include: one A/D and two D/A converters to produce digital input to and analog outputs from the computer, a Teletype for operator communication, a Connection Box to obtain flexibility in connections during the system setup, the Control Panel to place the necessary equipment (pushbuttons, amplifiers and relays), an X-Y plotter and the Analog Computer. A fairly basic block diagram is shown in Figure 5.1.

There are two PIAs used for this project apart from the PIA-3 used for interfacing the TTY.

Because it is possible to arrange the control lines CA1, CB1, CA2 and CB2 of a PIA on the peripheral side as inputs or outputs and because the MPU may control the conditions of these lines on request, the "status" line of the A/D converter was connected to the control line CB3 and the "Convert Command" line

was connected to the control line CB4. The necessary software was prepared to control and complete the conversion. The second side of the PIA-1 has been established as the output side and the 8-parallel data lines on the peripheral side of the PIA-1 were connected to the inputs of the converter D/A-1. The control lines CA1 and CA2 have been established by the software as input and output, respectively.

One side of the second PIA is also connected to another D/A converter named D/A-2. The control lines of this side CA3 and CA4 were established again by the software as input and output, respectively. Only one control line CB2 of the second side of the PIA-2 was established as input.

The commands SELECT, TEST, TRIP and PEN (on - off) were connected to CA1, CB2, CA3 and CA4, respectively. CA2 was used in connection with the relay R2 and R3 in order to obtain the computer control of the relays' operations.

The A/D and D/As consist of ZN425 8-bit dual mode single chip converters. The clock frequency of the A/D is 259.9 KHz. For full-scale reading the maximum conversion time is

$$\frac{2^8}{259.9 \times 10^3} = 0.985 \text{ millisecond}$$

The full scales of the converters have been set at 3.8 volts.

5.1.1 Connection Box

From both PIA-1 and PIA-2 six out of eight control lines CA1, CA2, CB1, CB2, CA3 and CA4 have been connected to 4 mm sockets on the box. Two D/A output lines and one A/D input have also been connected to three 4 mm sockets. In addition, the grounds of the converters have been placed on this box to obtain a reference for the input and outputs. Figure 5.2 illustrates the top view of the connection box.

This box has been prepared for quick and easy alteration of

the connections during the system setup.

5.1.2 Control Panel

A control panel with three pushbuttons was built. The buttons latch down and glow when pressed. They have to be pressed a second time to release them. Each button energises a circuit made up with two invertors connected in series and the output of each circuit is connected to a 4 mm socket. When the button is on there is 5 volts available at the associated socket. These pushbutton circuits provide adequate interfacing of the commands SELECT, TEST and TRIP to the PIAs' control bits. In addition, the TRIP button has been directly connected to a double pole form C D.I.L. Reed Relay (R1) to operate two relays in one package, simultaneously, Figure 5.3.

A Zener Diode Circuit with the gain 0.5 was built to interface the analog model output with the maximum 3.1 volts to the computer. The Zener breakdown voltage is 6.2 volts. Refer to Figure 5.4.

To operate two relays R2 and R3 by means of one control line (CA2) of the PIA, the circuit in Figure 5.5 was established. The point 5 is a 4 mm socket.

The box also houses three non-inverting OP Amps used to secure no-current-flow to the plotter and analog computer. One of the purposes of the OP Amps is to effect the calibrations. Refer to Figure 5.6.

5.1.3 Hardware System Setup

The overall diagram is shown in Figure 5.8.

The test system was connected to the mathematical model of the Governor. Once the Analog Computer is switched to the "compute" mode, all the test sequencing is achieved from the Control Panel, PTY, and the Microcomputer.

A sheet of A4 chart paper is used to obtain a permanent record of the valve performance. How such a paper has been arranged is explained in the System Software part of the thesis.

When the software is run, the SELECT button has to be pressed to start the step-by-step testing sequence. When this button is on, the plotter identifies the required valve by underlining one of the pre-printed code numbers on the plotter paper. Then a constant input is fed to the model. The TEST button is pressed to close the valve by decreasing the Governor input to zero. Then the TRIP button is pressed to trip the Power Piston. Releasing the TEST initiates a command to increase the Governor input signal from nought to its pre-test value. Then the TEST button is pressed to decrease the Governor input to zero. At the bottom stop of the plotter pen the TRIP and TEST buttons are released in sequence. When the valve regains its pre-test position the SELECT button is released.

Driver

The Governor signal obtained from the computer is sent to the converter D/A-2 and the analog signal at the output of the D/A-2 is fed to the OP Amp-1. The gain of the amplifier is 2.26. The purpose of the amplifier is not only to prevent the converter from the possible current-sink effect of the plotter and the Analog Computer, but also to amplify the plotter's Y input signal during the valve identification procedure. The amplifier output is applied to the Y input of the plotter and to the pot 2P13 whose value is 0.306 when the relay R2 is short-circuited. This pot is on the Analog Computer. A summing amplifier (2A3) of the Analog Computer is utilised to obtain the gain of 10 and

sign inversion. The sign inversion was necessary since the negative Governor input gives a positive Governor output.

The values of the OP Amp-1, the pot 2P13 and the amplifier 2A3 were arranged so that the amplitude of the Governor input would be 26.25 volts for the maximum D/A-1 output 3.8 volts. The obtainable maximum gain of the summing amplifier is not more than 10.

Sensor

The Zener Diode Circuit receives the Position Output, multiplies it by 0.5 and feeds to the OP Amp-3. The purpose of the circuit is that the diode regulates the A/D input against variations in the model output and also against variations in the A/D input current. The Zener breakdown voltage is 3.6 volts. Since the maximum value of the Position Output is less than 7.2 volts, the gain of the circuit was chosen to be 0.5. The OP Amp-3 with the gain of 1 was placed to prevent the interaction between the Zener Diode Circuit and the A/D converter. The output of the OP Amp-3 supplies the signal to the A/D and, when the relay R3 positioned to the X input of the plotter over the OP Amp-2 whose gain is 3.35. The OP Amp-2 prevents the current draw of the plotter from the converter D/A-1. The gain was chosen higher than one in order to obtain more distinct record of the valve performance and large movement on the X axis during the test sequencing.

Other Connections

The outputs of the SELECT, TEST and TRIP circuits were connected to CA1, CB2 and CA3 sockets on the Connection Box over the specially prepared leads shown in Figure 5.7. The reason why the leads were arranged was the fact that the PIA control lines CA1, CB2 and CA3 were floating when their inputs (SELECT, TEST and

TRIP) were zero. With this arrangement of connections it was ensured that when their inputs are zero these control lines are earthed and do not float.

The relays R11 and R12 in the relay package R1 were arranged as in Figure 5.8. When TRIP is on the input of the pot P2 is earthed and the input of the amplifier B4 over $0.1\text{ M}\Omega$ is connected to its output. This R11 operation produces the same effect as the Trip Test Relay does on the actual valve. Although the pot P2 input was earthed the amplifier B4 would still integrate and produce non-zero output which is partly the result of the fact that the gain of B4 is as big as 100. For this reason the relay R12 was felt to be necessary. In fact, even if the B4 input over $0.1\text{ M}\Omega$ was earthed when TRIP was on, the amplifier would still integrate. That was why the B4 output would be short-circuited with this B4 input when TRIP was on.

The CA4 socket on the Connection Box was connected to the PEN command input of the plotter over a piece of wire. The CA4 is set to +5 V or to 0 V through the software. When the PEN command input is supplied with +5 V the pen is on, otherwise it is released.

The socket called CA2 is connected, over a wire, to the point 5 socket on the Control Panel that positions the relays R2 and R3. The socket CA2 is energised and de-energised under the software control. This enables the computer to control the operations of the relays R2 and R3 when the socket 5 is connected to CA2. The D/A-1 output is used to give the X movement during valve identification on the plotter. This computer output is not utilised before or after the valve identification.

5.2 Software Development

The software has been prepared as program packages for ease^{of} applications-oriented non-programmers. The system software includes 10 subroutines whose purposes in general are: creating the negative and positive going ramp signals at one of the microcomputer's output ports, reading the present input signal at the input port, storing the data used for the valve identification in the dedicated memory locations, and enabling the TTY to print the commands that inform the operator of what to do next. [15].

5.2.1 Pletter Paper Arrangement

There are 20 different numbers on the actual pletter paper sheet in two columns each consisting of 10 numbers. Each number belongs to one valve on the system.

The type of paper arranged to be used in the microprocessor-based control system is illustrated in Figure 59. The distance between two subsequent numbers on the same column has been assumed to be 1.5 cm and the distance between two numbers on the same row has been assumed to be 2 cm. Also it was assumed that the pen of the pletter draws 1 cm length of line under the number during the valve identification of the pletter. For these reasons the scaling on both axes has been taken to be 500 mV/cm.

The gain on the X-axis is 3.35 and on the Y-axis is 2.26.

To under-line one of the numbers in the first column the X-axis should start with the decimal 225 and finish with the

decimal number 235 and the Y-axis should be adjusted to one of the following decimal numbers according to the valve numbers: 24, 46, 68, 90, 112, 134, 156, 178, 200, 222.

The decimal values of Y above are still valid for the following second column, but the starting and ending values on the X-axis should be decimal 245 and 255, for the numbers on the second column.

Table 3 summarises the values on both X and Y-axis required to underline each valve number. (See the example in APPENDIX C)

5.2.2 Preparation of The Ramp Signal

During on-load valve test sequencing, the actual steam valve movement from the fully open to the fully closed position takes approximately 3 seconds. Since the step input response of our Governor model is ten times slower than the actual system response, this ramp speed has also been assumed to be, 30 sec, ten times slower than the actual 3 sec duration.

Since the 8-bit converters are used in the project, it was assumed that the maximum opening of the valve will occur when the full scale reading FF of the converter in the driver circuit exists.

The hexadecimal number FF equals to 255 decimal. To enable the computer to produce a positive going ramp signal, nought to maximum, we may write a program where an accumulator is increased by one and each time the new value of the accumulator is stored at the output. Since this will be a loop repeated

Decimal Number to be put on		VALVE NUMBERS ON A P L O T T E R P A P E R																			
		First Column									Second Column										
		1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
X-axis	Starting	225									245										
	Finishing	235									255										
Y-axis		222	200	178	156	134	112	90	68	46	24	222	200	178	156	134	112	90	68	46	24

Table 3.

255 times because before the start the accumulator had been cleared, and it is required that the loop should take 30 sec, the time spent between the output two subsequent accumulator contents must be: $\frac{30 \text{ sec}}{255} = 0.117647058 \text{ sec}$.

The first four figures after the comma may give the adequate approximation. But the value is being taken as it is to give some idea of how the command NOP (NO Operation) may be used for accuracy.

Normally the following piece of program is enough to store the content of the accumulator B after identifying the output address and clearing B:

```

        POSGO  STAB  dddd
                INCB
                CMPB  mmmm
                BLS   POSGO
    
```

In this program dddd is the output address and mmmm is the address of the memory location whose content is hexadecimal FE (decimal 254). This program takes $5+2+4+4=15$ cycles and, because the MPU used in this work executes one cycle in $2 \mu\text{sec}$, $15 \times 2 = 30 \mu\text{sec}$

$$117.647058 \text{ msec} \approx 0.030 \text{ msec}$$

Now a delay loop should be added. The delay loop is:

```

        LDX  kkkk
X1     DEX
        BNE  X1
    
```

Here kkkk is the content of the index register. LDX kkkk takes 3 machine cycles or $3 \times 2 = 6 \mu\text{sec}$ of execution time.

$$30 \mu\text{sec} + 6 \mu\text{sec} = 36 \mu\text{sec} = 0.036 \text{ msec}$$

Then, we find that

$$117.647058\text{msec} - 0.036 \text{ msec} = 117.611058 \text{ msec.}$$

DEX and BNE X1 together take $4+4 = 8$ cycles or $16 \mu\text{sec} = 0.016\text{msec}$

The content of the index register then should be

$$\frac{117.611058\text{msec}}{0.016 \text{ msec}} = 7350.691125$$

The decimal 7350 equals to the hexadecimal 1CB6.

Since

$$7350 \times 0.016 \text{ msec} = 117.6 \text{ msec,}$$

there is still need for a delay time of $11.058 \mu\text{sec}$

because

$$117.611058 \text{ msec} - 117.6 \text{ msec} = 0.011058 \text{ msec} = 11.058 \mu\text{sec}$$

Since one NOP takes 2 cycles or $4 \mu\text{sec}$, two or three NOP should be added. Two NOPs take $2 \times 4 \mu\text{sec} = 8 \mu\text{sec}$ and three NOPs take $3 \times 4 \mu\text{sec} = 12 \mu\text{sec}$.

Since $12 \mu\text{sec}$ is nearer to $11.058 \mu\text{sec}$ than $8 \mu\text{sec}$ is, it has been thought that adding 3 NOPs would be adequate.

This way of thinking made the loop POSGO to be:

PGOING	STAB	dddd
	LIX	1CB6
X1	DEX	
	BNE	X1
	NOP	
	NOP	
	NOP	
	INCB	
	CMPE	mmmm
	BLS	PGOING

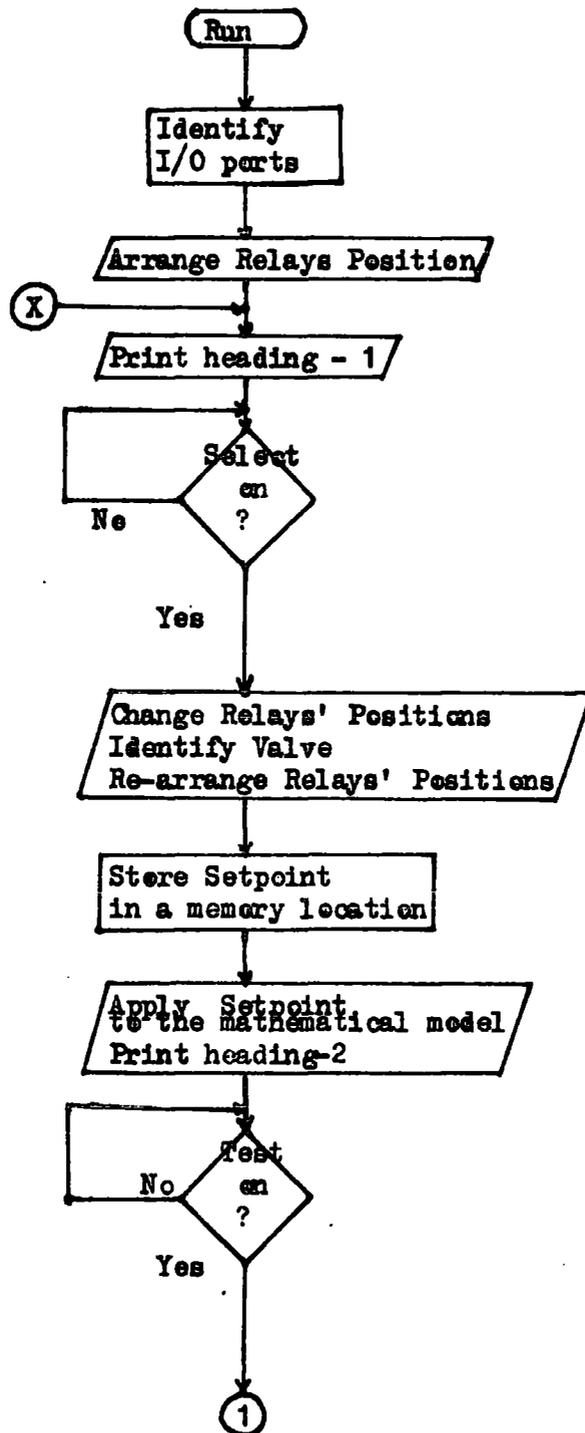
The same idea is used for the sub-program that would give the negative going ramp signal. The sub-program is:

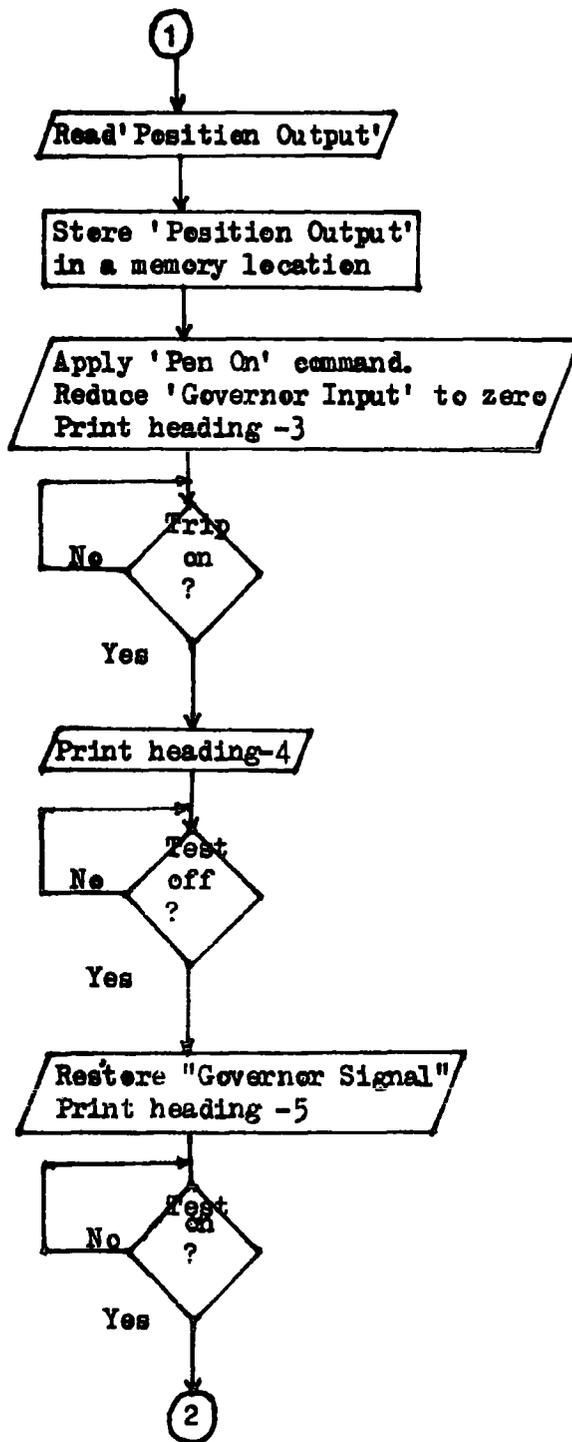
```
      NGOING STAA  dddd
L9      LEX   1CB6
        BNE   L9
        NOP
        NOP
        NOP
        NOP
        NOP
        NOP
        DECA
        BNE  NGOING
```

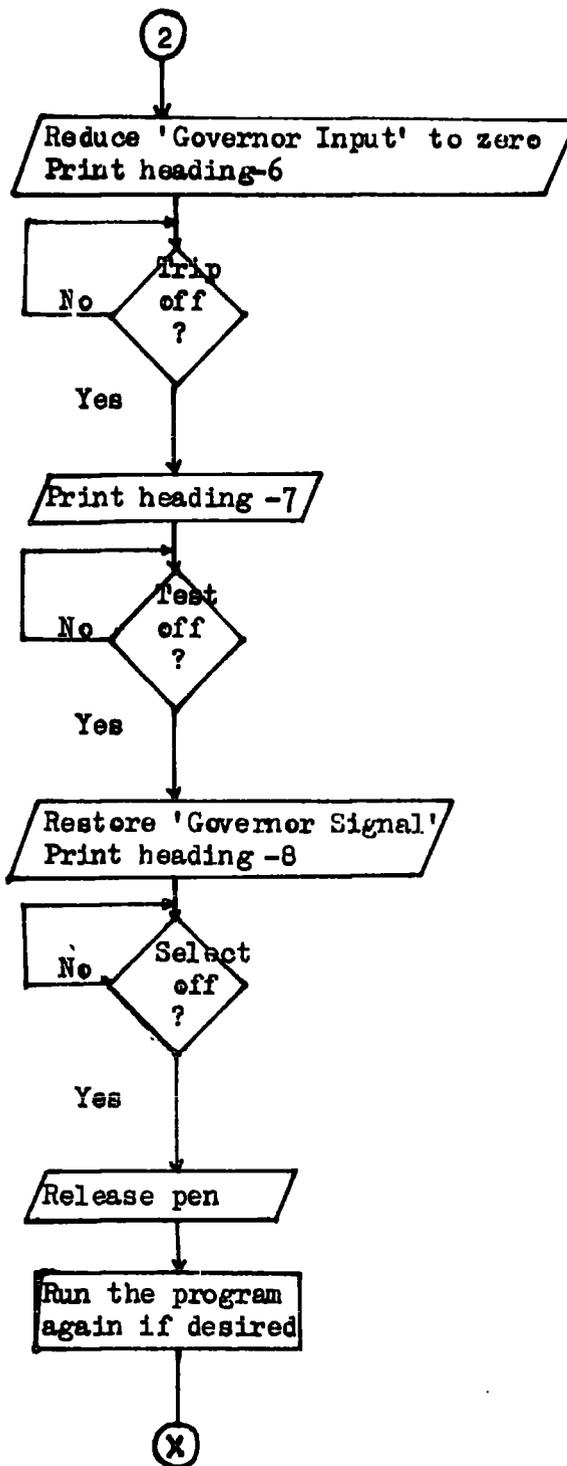
Here two more NOPs have been added to comprise the same length of time that CMPB mmmm in the previous sub-program spends.

5.2.3 Flowchart

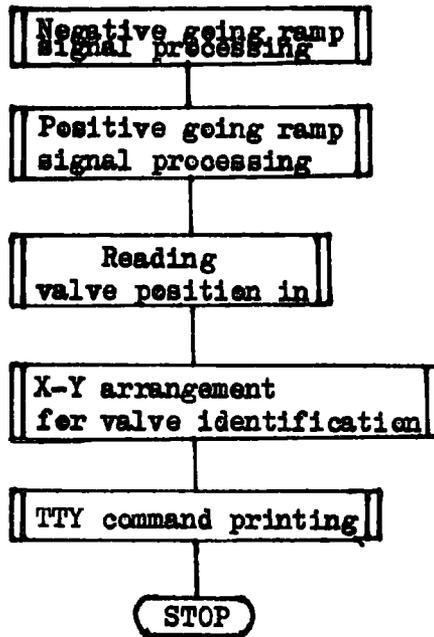
The following flowchart was prepared to organize the problem.







The system subroutines;



5.3 On-Load Valve Testing Procedure

When the program starts running the relay R2 is short-circuited, the relay R3 has its points B and 2 short-circuited. Refer to Figure 5.8. The pen of the plotter becomes off.

The TTY prints:

```
( 1 ) PRESS SELECT PUSHBUTTON  
(
```

The MFU checks in a loop if the SELECT is on.

Select Valve

As the SELECT command^{is} reached, the relays R2 and R3 change their positions so that the model on the analog computer gets no input signal and the X axis of the plotter gets switched over from its valve position input to the output of the converter D/A - 1. The computer generates the X and Y movements over the D/A -1 and D/A -2 respectively. The PEN ON command is also created by the software and fed to the pen circuit of the plotter. The Y input value is kept constant while the X input is being increased. After underlining the valve number on the plotter, the pen is released and it goes to its set-point since the inputs to both plotter axes are zeroed by the program. Then the relay R2 is short-circuited and the X axes is switched back to the valve position input. After this is done, a constant input is fed to the model and the Y input of the plotter over the OP Amp -1. Obviously the X input gets a position output as a response to the input signal fed to the mathematical model. The value applied as the input to the model is also stored in a memory location.

The TTY prints: 2) PRESS TEST PUSHBUTTON

(

Then the MPU starts checking in a loop if the TEST ON command has been applied.

Press Test Pushbutton

When TEST is on, the computer reads the valve position and stores it in a memory location. Then the PEN ON command is applied and the stored value on the D/A -2 is decreased to zero. This results in zero output of the valve position. The plotter draws the output of the model on the analog computer, which is the valve position, against the reducing input, continuously. This is the line \overline{ABO} shown in Figure 5.11.

After the pen has reached at its bottom step, the TTY prints: 3) PRESS TRIP PUSHBUTTON

(

The MPU starts checking in a loop if TRIP is ON.

Press Trip Test Pushbutton

When the TRIP button is switched on the relays R11 and R12 change position so that the input of the pot P2 is earthed and the input of the amplifier B4 is switched from the output of pot P12 to its own output. (Refer to Figure 5.8). The TTY prints: 4) RELEASE TEST PUSHBUTTON

(

Then the MPU checks in a loop if the TEST button has been released.

Release Test Pushbutton

This initiates the execution of a subroutine which increases the input voltage to the model on the analog computer. This input value is increased from zero to the value existing before the first TEST ON command was applied. Since the TRIP ON command has switched off the Power Piston input, the X input of the plotter stays at zero and the Y input moves from zero to the point C. Figure 5.11 shows this by drawing a line between the points O and C. Then the TTY prints: 5) PRESS TEST PUSHBUTTON

(

The MPU checks in a loop if the TEST button has been pressed. The pen stays at the point C in Figure 5.11 until the TEST button is pressed down.

Press Test Pushbutton

This starts the execution of a subroutine that would give the decrement on the input value of the model until this input signal reaches at zero. During the decrement of the model input the pen moves from the point C to O indicating that the TRIP is still on and the Governor is getting the closure signal. When the pen reaches at its bottom step, the TTY prints:

6) RELEASE TRIP PUSHBUTTON

(

The MPU starts checking in a loop if the TRIP button has been released.

Release Trip Test Pushbutton

Releasing the TRIP button results in the position changes of both relays R11 and R12 simultaneously so that the input of the pen P2 is switched back to the output of the amplifier 2A2 and the input of the amplifier B4 is switched back to the output of the pen P12. (Refer to Figure 5.8). Then the TTY prints:

7) RELEASE TEST PUSHBUTTON

(

The MPU checks in a loop if the TEST button has been released.

Release Test Pushbutton

The Governor gets its operating value, before the test sequencing had been started by pressing the SELECT button back. The pen moves from its bottom-stop to the point A over the line ODA in Figure 5.8. The the TTY prints:

8) RELEASE SELECT PUSHBUTTON

(

The MPU checks in a loop if the SELECT button has been released.

Release Selector Pushbutton

This initiates a command to release the pen.

5.4 Test Results

For a typical test, the program was run after having adjusted the specified setpoint of the plotter. At the end of the test, the teletype commands shown in Figure 5.10 and the plotter responses illustrated in Figure 5.11 were obtained.

As it is seen in Figure 5.11, there is a phase difference between the valve closure line \overline{ABO} and the valve opening line \overline{ODA} which occurs because of the overall time constant in the analog Governor. If the delay did not occur during the closing and re-opening of the valve, the pen movement between A and O would be a straight line. Added to this there is a further delay which results from the processing time of the computer and the D/A - 2's conversion time. Since storing the contents of an accumulator in extended addressing mode takes five machine cycles and one machine cycle takes $2\ \mu\text{sec}$ of processing time, loading data on a D/A converter takes $10\ \mu\text{sec}$. The conversion time of the D/A for full scale reading is 1 millisecond.

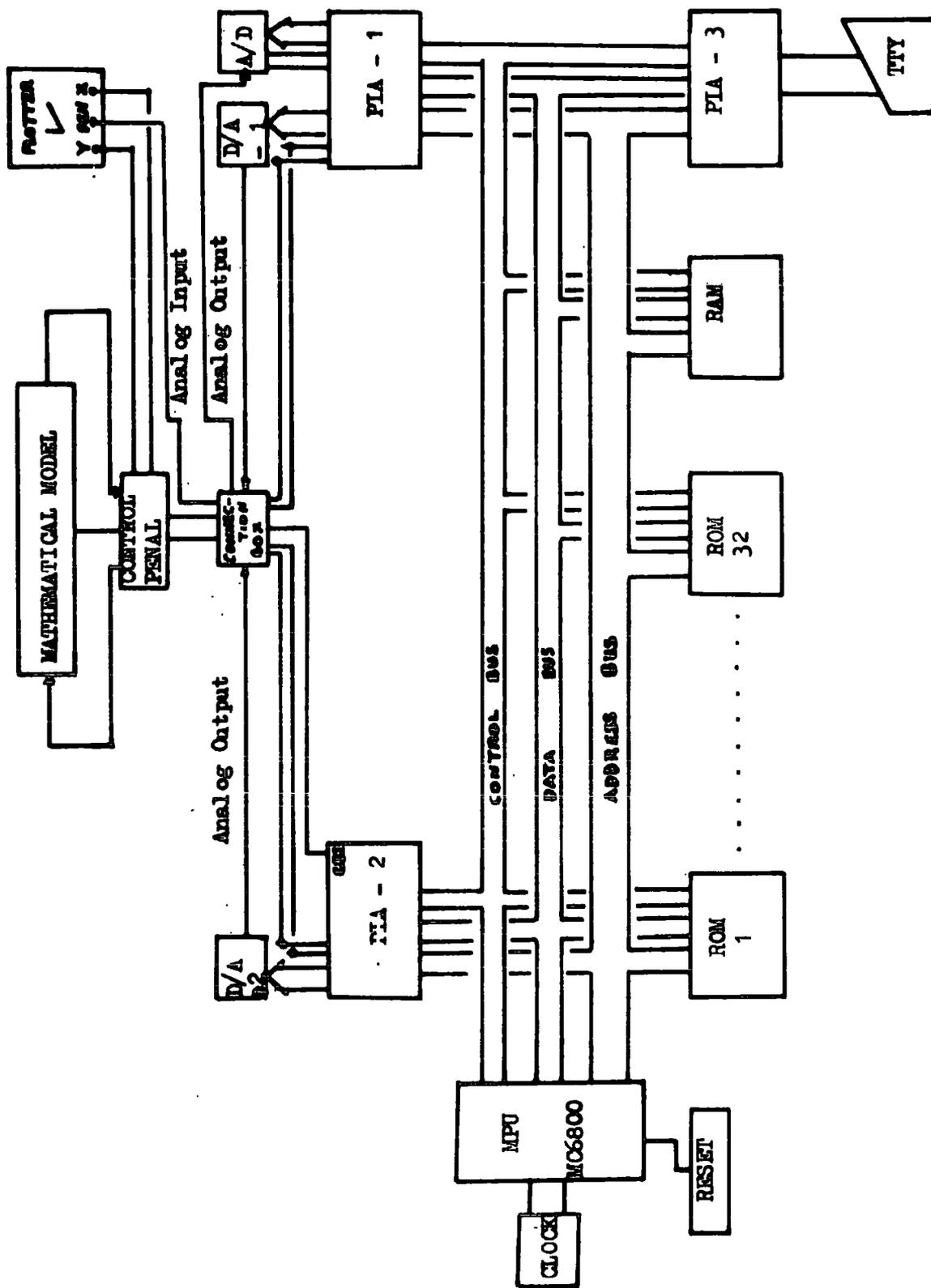


Figure 5.1 Overall system block diagram

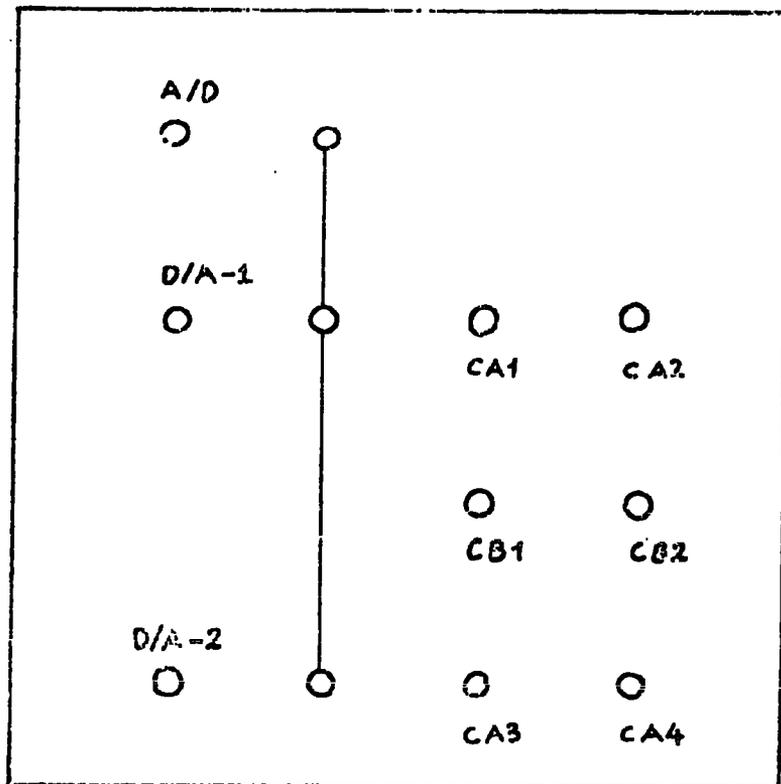


Figure 5.2 The top view of the connection box.

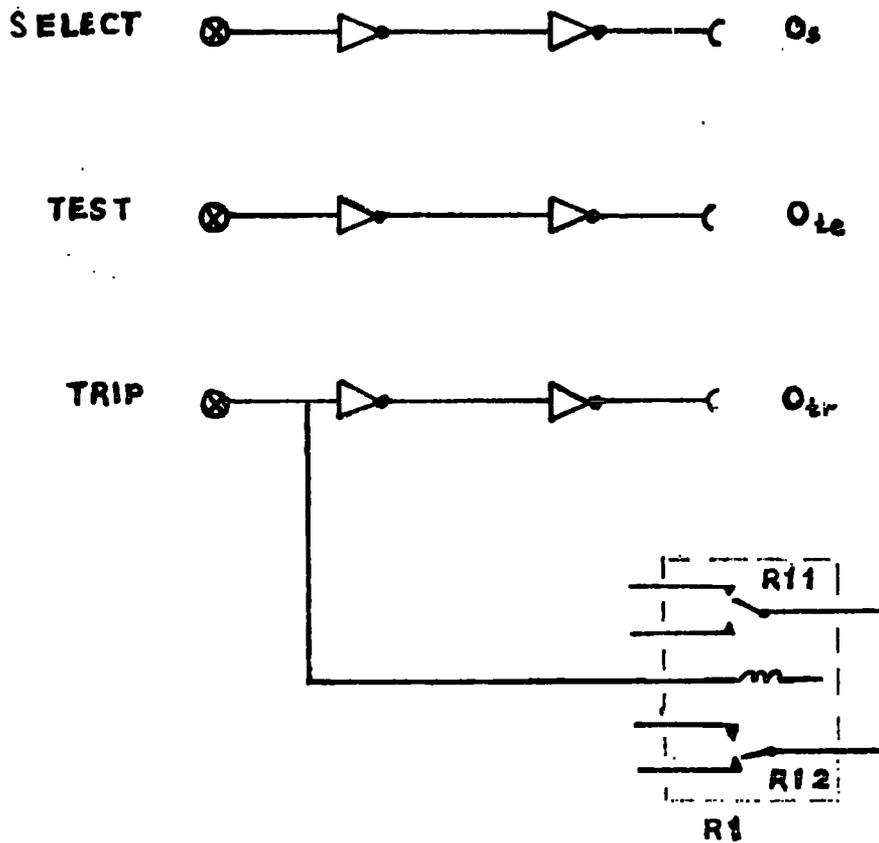


Figure 5.3 Pushbutton connections.

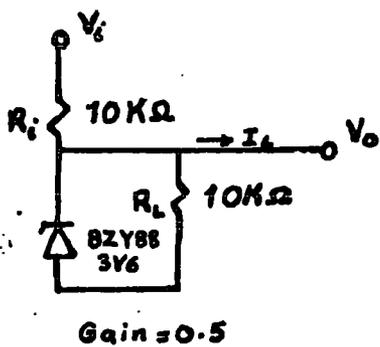


Figure 5.4 Zener diode circuit

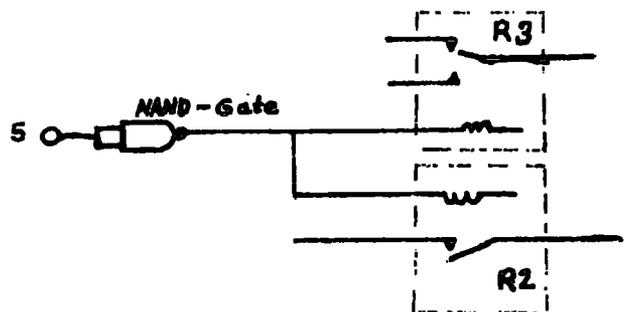


Figure 5.5 The operating input connections of the relays R2 and R3.

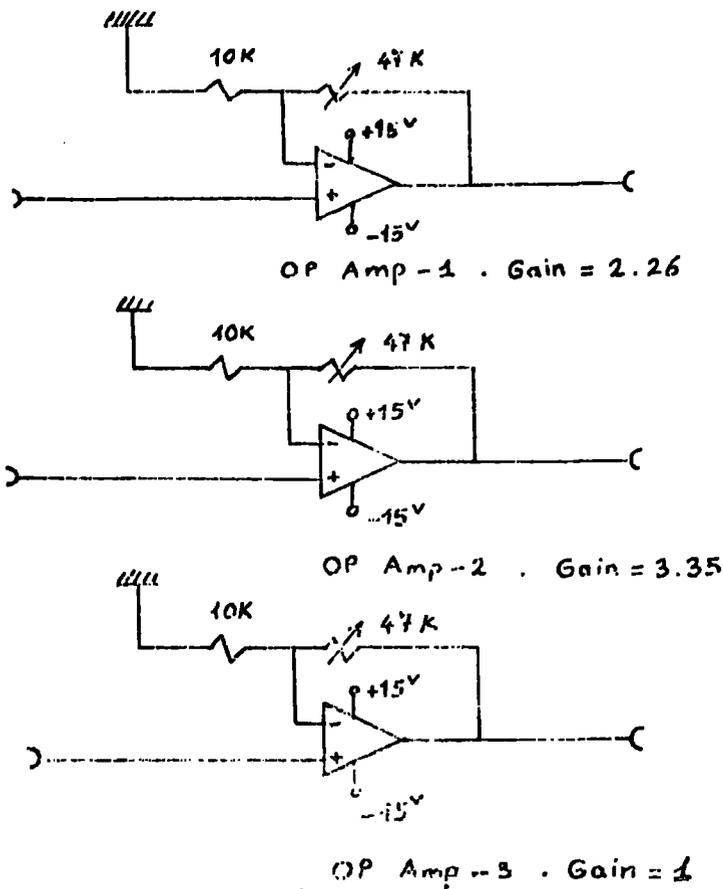


Figure 5.6 Existing OP Amp connections in the control panel.

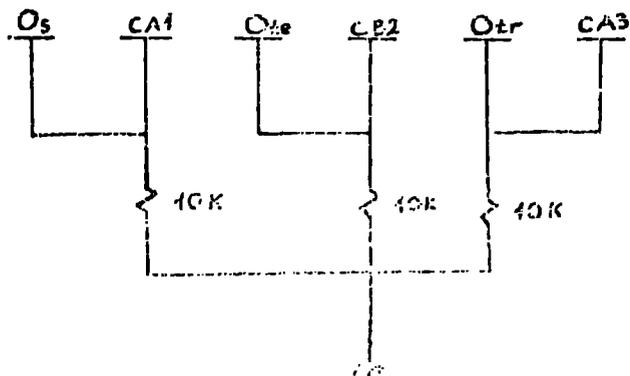


Figure 5.7 Special purpose leads

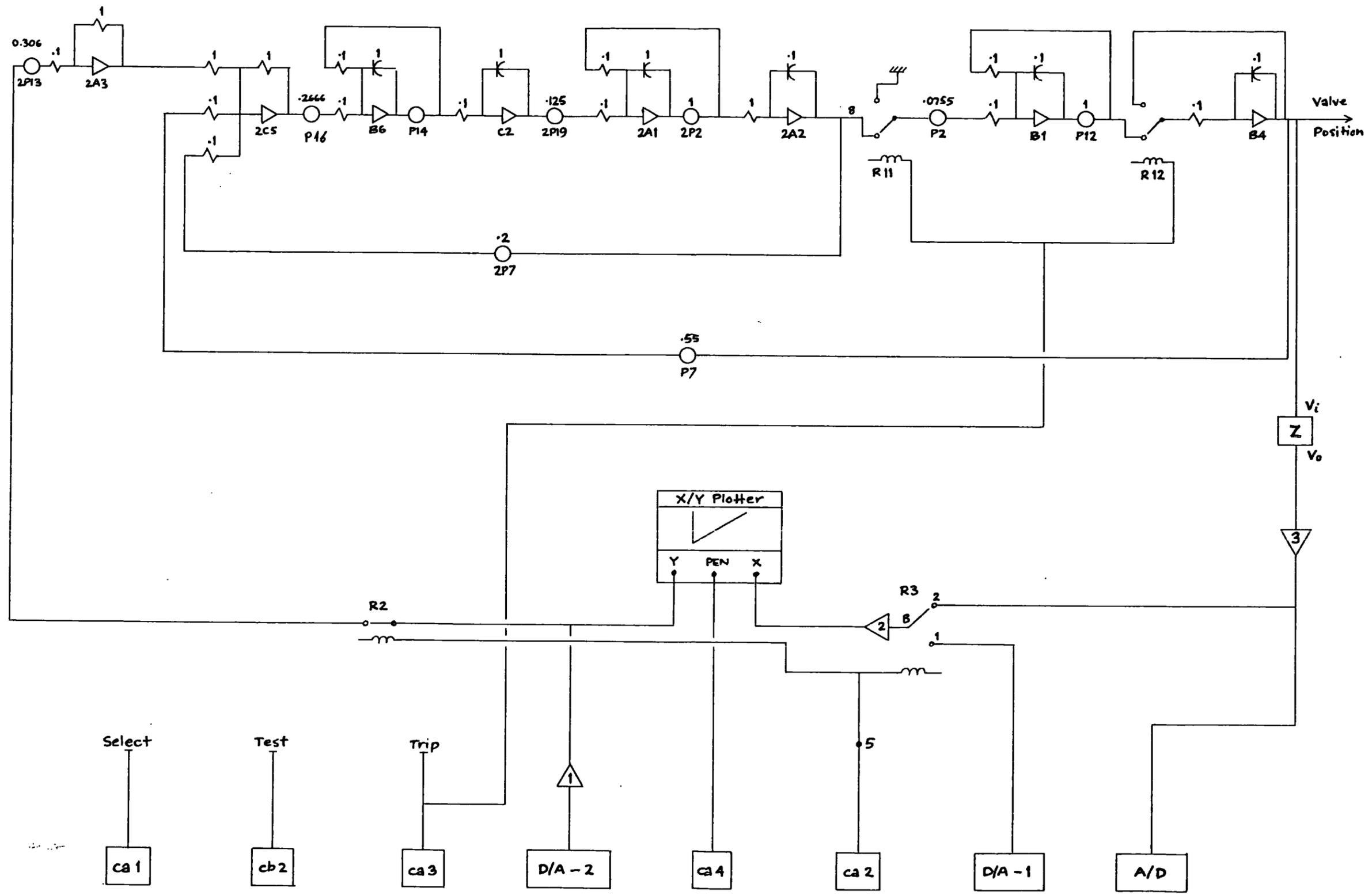


Figure 5.8 Valve testing system diagram

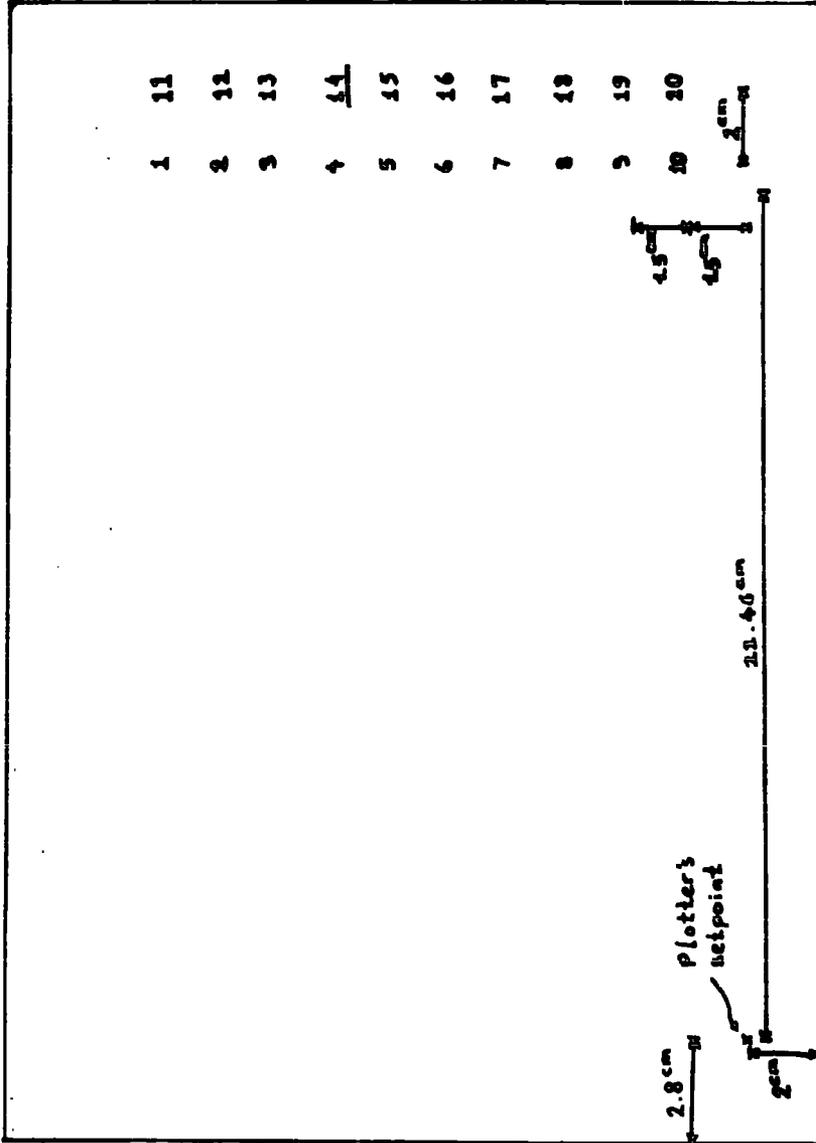


Figure 5.9 Dimensions on A4 plotter paper used for microprocessor-based on-load valve testing.

- (1) PRESS SELECT PUSHBUTTON
- (2) PRESS TEST PUSHBUTTON
- (3) PRESS TRIP PUSHBUTTON
- (4) RELEASE TEST PUSHBUTTON
- (5) PRESS TEST PUSHBUTTON
- (6) RELEASE TRIP PUSHBUTTON
- (7) RELEASE TEST PUSHBUTTON
- (8) RELEASE SELECT PUSHBUTTON
- (

Figure 5.10 Teletype commands of the microprocessor -based on-lead valve testing obtained at the end of the test procedure.

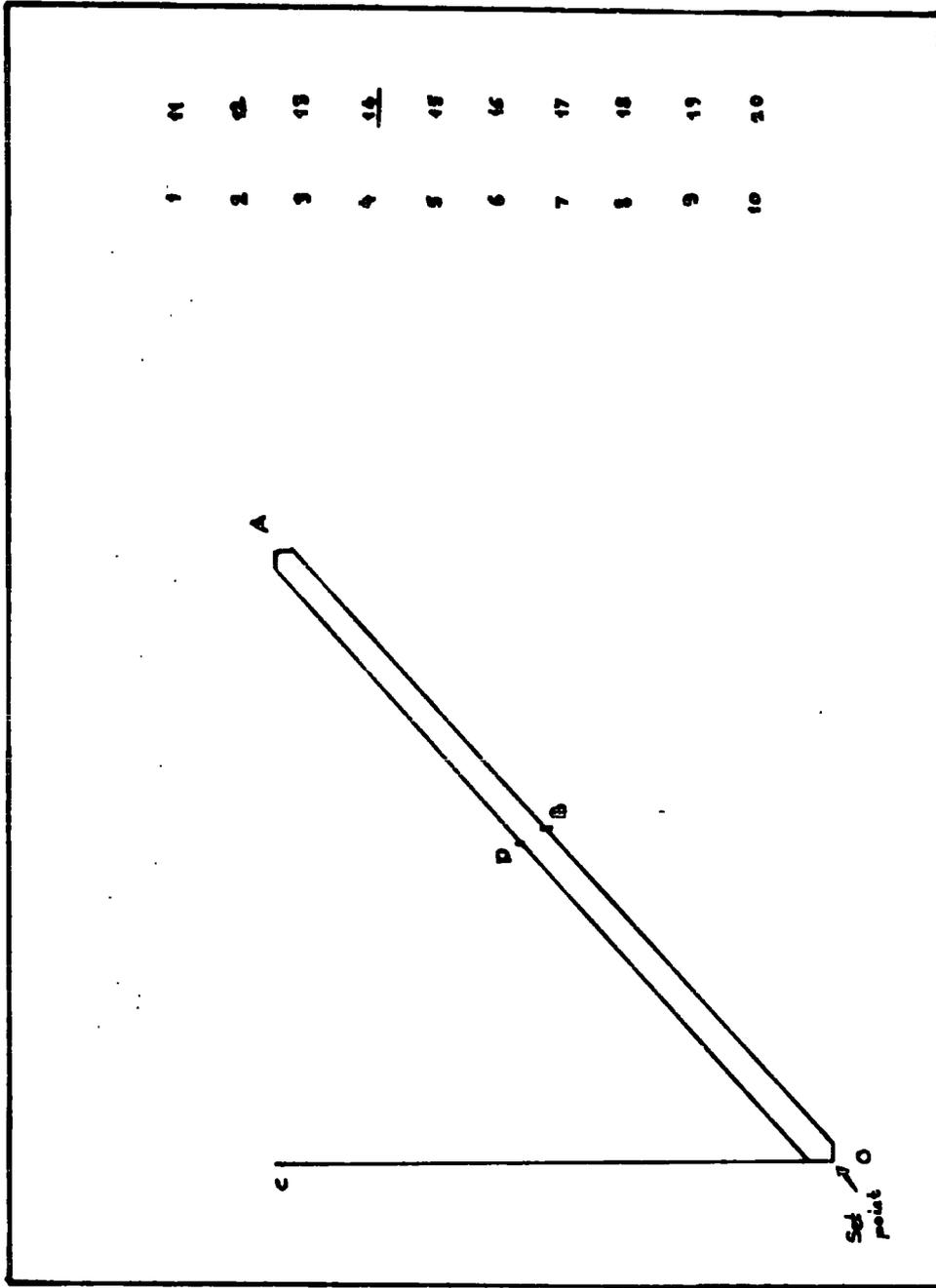


Figure 5.11 The plotter record of the microprocessor-based on Lead Valve testing system obtained at the end of the test procedure.

CHAPTER VI

MICROCOMPUTER IMPLEMENTATION OF THE EXISTING TESTING SYSTEM

This chapter deals with the differences between the existing analog on-load valve sequencing and its microcomputer implementation. It also discusses the different ways of how the implementation can be adapted on the actual system. Figure 5.8 should be referred ^{to} along ^{with} this chapter.

Some discussions about the system software have been found to be necessary and they are placed in this chapter. Since the on-load valve testing of one valve has been achieved in the project, a basic idea of how the microprocessor-based testing system may be utilized to test a number of valves of this type is also given.

6.1 Differences of Two on-Load Valve Sequencings

When the Select is on, the actual system identifies the valve by encircling the valve number on the left-hand side of the plotter paper, but the model draws a line under the number on the right-hand side of the paper.

Since the pushbuttons used in this project glow when pressed there is no need of any other signal indicating "On-Load Test" in progress.

The procedure of checking the majority-voting circuit

in the Serve-Value Amplifier during the ramp operation may be achieved with additional hardware circuits and extra software for this purpose. The software then may be placed with the delay loops in the subroutines NGOING and PGOING of our program. In fact the execution of the software for this purpose is time-limited. This limit is 117.617 msec for our program where the ramp takes 30 sec to make the valve move from the fully open to the fully closed position. See CHAPTER V, 5.2.2.

If the relay package R1 is disconnected from the Trip pushbutton and operated by a computer controlled separate signal, like the relays R2 and R3, there will be no need to use extra switches as interlocks, to prevent the accidental valve movement resulting from inadvertently depressing the Trip button. When the Trip is on, however, the computer controlled relay package R1 will receive a signal over an extra output channel from the microcomputer, and consequently the Power Piston will become tripped. This tripping position of the relay will be kept until the Trip is switched off at the end of the test. When the Trip is checked and found to be off by the computer, the program will re-arrange the position of the relays R11 and R12 in the R1 package.

The Governor and the Test signals in our case have been assumed to be produced by the computer and fed in the same input. In fact, on the actual system, the test signal which is a ramp signal is added to the Governor signal and sub-

tracted from the Governor signal during the valve test. A method by which this sort of operation can be simulated is now discussed.

6.2 The Procedure of The Test Signal (Ramp)

Addition To Governor Signal

On the actual system, the ramp signal is added to and subtracted from the operating Governor signal. In this project, however, it was assumed that even the Governor signal was being obtained from the computer. Because of this assumption, after the valve selection had been completed a constant voltage created through the software was fed into the Governor Model, and it was zeroed to close the valve when the Test button was on and it was increased back to its operating value when the Test button was off. This can be arranged so that the ramp signal may only be added to a constant signal supposed to be the Governor signal and then fed to the Governor. Then the hardware and software should be rearranged. The changes in the hardware configuration in Figure 5.8 are shown in Figure 6.1. Here, the summing amplifier O1, the single inverter O2 and the potentiometer 12PX are on the Analog computer. The re-arranged hardware also includes a double-pole relay R4. The purpose of the pot 12PX is the calibration. When the Select pushbutton is on, the computer changes the position of the relays R2, R3 and R4 in Figure 6.1. These positions' change ensures the fact that during the valve identification the Governor is still on-load, and the X and Y movements of the platter generated from the microcomputer are

not effecting the operating Governor signal. After the identification procedure the relays mentioned above re-gain their positions in Figure 6.1. At this stage of the software already written there is no need of a constant input from the microcomputer to the OP Amp-1 in Figure 5.8. to make the valve on-load since there is a positive input called "INPUT" to the summing amplifier O1 in Figure 6.1. To provide the close Governor signal the amplitude of the computer generated ramp signal at the output of the amplifier 2A3 should be increased to be equal to the constant input signal. This can be done either by reading the constant input in and letting the microcomputer know how high the ramp amplitude is going to be or by reading the error and checking the error against the ramp signal during the time the ramp is being generated. When the error becomes zero the ramp amplitude must not be further increased. Alternatively, the restoration of the operating Governor signal may be achieved by decreasing the generated ramp to zero.

In the system software, the positive and negative ramp generating subroutines, PGOING and NGOING respectively, would be exchanged.

6.3 Discussion Based On The System Software

In the supervisory part of the software the Interrupt Request (IRQ) flags of the PIAs Control Registers were cleared before the control modes were set, and the Data Direction Registers of the output ports were cleared before the identification of these parts as the outputs. For this reason only

twenty-six memory locations were needed. In fact, pressing the RESET button on the computer before running a program sets all PIA registers to zero so that these memory locations might be saved. But the flag clearing in this section of the firmware guarantees the fact that no malfunction will occur because of the software even if a flag had been set for any reason before the program was run.

For the teletype (TTY) commands each word to be printed has been written in subroutine-form and an adequate set of these subroutines are called each time the operator's attention is to be drawn. This way of printing the commands calls for less than four hundred memory locations (356). However, if each piece of the program of the commands to be printed on the TTY was placed at the beginning of each system test step, 1140 memory locations would be needed. By preparing the software portion for this purpose in subroutine-forms more than 700 memory locations have been saved.

6.4 Adaptation of the Microprocessor-Based Single Valve Testing System To Multi-Valve Testing System.

Only one Governor System controlling the position of one valve has been simulated and valve testing of one unit has been outlined in this project.

To adapt the valve testing system in Figure 5.8 to a multi-valve testing system some hardware and software reconfigurations must be done. In the single valve testing system it was assumed that the operating Governor signal was processed by the computer, and to close the valve this signal was zeroed and then increased

to open the valve. If all the valves on a steam turbine system are receiving their operating signals from the computer output channels, then closing or opening each valve requires the signal on each dedicated analog output to be zeroed or increased by means of the system software. If one select pushbutton is dedicated to each valve, all the select signals may be input to the computer. The software prepared for this purpose checks each select input in turn and when one of the select buttons is on then the program loads the necessary data in the dedicated memory locations to be referred during valve identification. The computer may also switch an X-Y plotter's X input from one valve position output to a computer output channel and then provide both X and Y movement. After a valve identification the computer switches the X and Y inputs of the plotter to the position output and the Governor input channel of the valve to be tested by arranging a set of relays. Since the computer will know which valve is to be tested after checking the select inputs, it should not be difficult for the computer to arrange the relay positions so that the plotter is switched to the valve under test. In the software, after establishing the I/O ports and the TTY has printed the "PRESS SELECT" command, the computer starts checking all select inputs in a loop. When one particular select button is on, then the program may jump to a subroutine and load the data dedicated to that particular valve in the dedicated memory locations to be referred to during the valve identification. In our program, the subroutine VALNUM stands for this purpose. If there are, say, twenty valves and select input channels, then there should be twenty subrouti-

nes each of which loads different maximum X, minimum X and Y values in the dedicated memory locations. When the Test is on, the position output of the desired valve may be switched over to one input channel of the computer. So there is a need of only one input channel to read the desired valve position. One test input will be enough to check if the test button has been pressed. When the valve closure or opening is desired, the software may call a subroutine to identify which computer output channel should receive the closure or opening signal. The common negative or positive going ramp processing subroutine may be used for any valve.

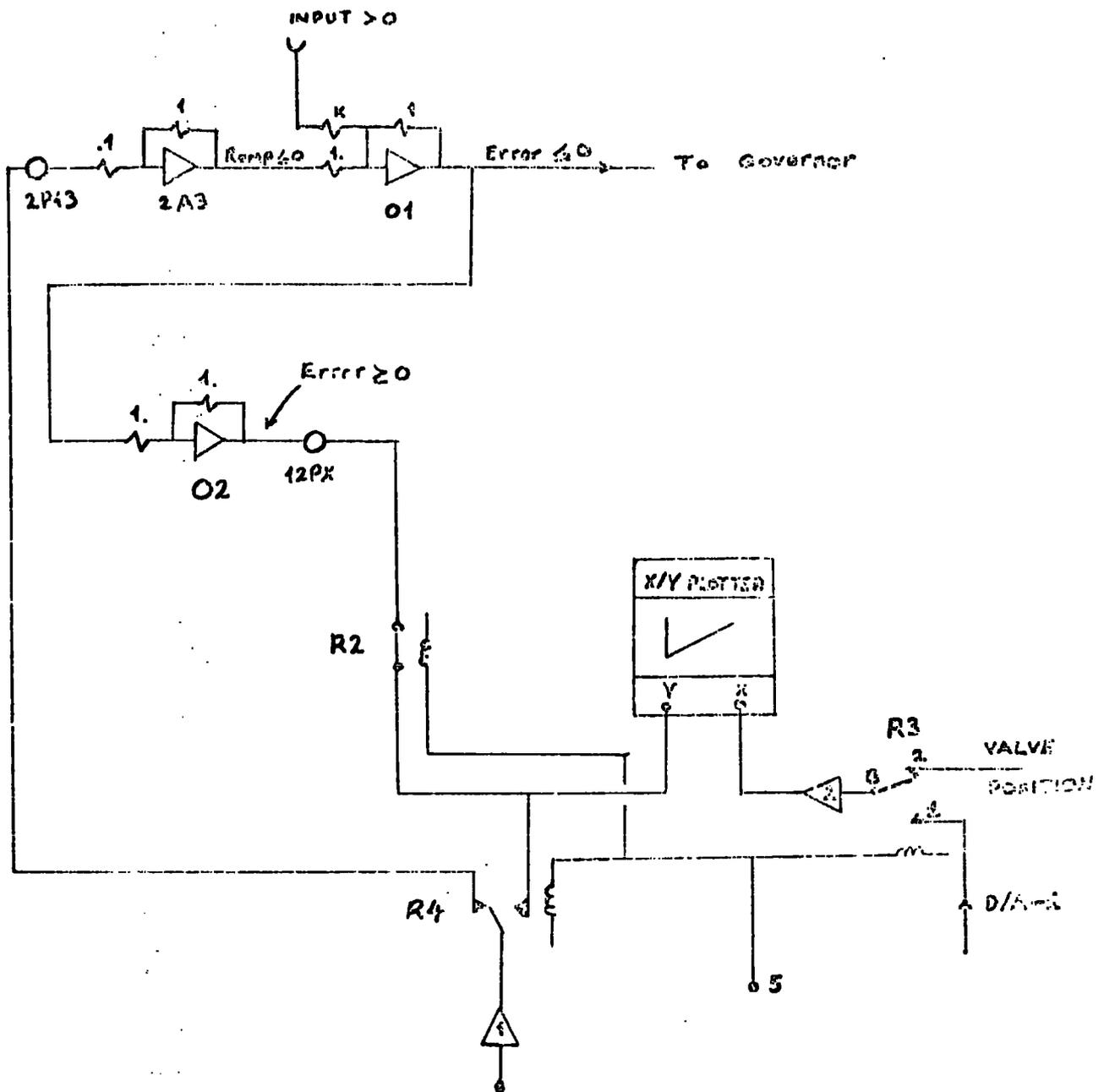


Figure 6.1 Hardware configuration for the simulation of the ramp signal addition procedure.

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APPENDIX A

BLOCK CALCULATIONS FOR ANALOG COMPUTER MODELLING

The block calculations of the reduced system have been done step by step as the following : [23] (E_i is block input signal and E_o is block output signal).

BLOCK I - Servo-Valve

$$\frac{G_2}{1 + T_5S}$$

$$\frac{E_o}{E_i} = \frac{G_2}{1 + T_5S}$$

$$E_o + T_5SE_o = G_2E_i$$

$$T_5SE_o = G_2E_i - E_o$$

$$T_5E_o = \int (G_2E_i - E_o) dt$$

$$\text{For } G_2 = 2.666 \text{ and } T_5 = 0.1,$$

$$\text{it can be written } 0.1E_o = \int (0.2666 \times 10 \times E_i - E_o) dt$$

Figure 7. (a).

BLOCK II - Primary Actuator

$$\frac{G_3}{(1 + T_6S)S}$$

$$\frac{E_o}{E_i} = \frac{G_3}{(1 + T_6S)S}$$

$$SE_o + T_6S^2E_o = G_3E_i$$

$$T_6SE_o = \frac{1}{S} \times G_3E_i - E_o$$

We may also write this equation in the following form;

$$T_6SE_o = \frac{1}{S} \times G_3 \times E_i - \frac{SE_o}{S} = \frac{1}{S} (G_3 \times E_i - SE_o)$$

In ^{the} time-domain, this becomes

$$T_6 \frac{dE_o}{dt} = \int (G_3 \times E_i - \frac{dE_o}{dt}) dt$$

For $G_3 = 1.25$ and $T_6 = 0.1$, the equation may be written as

$$0.1 \frac{dE_o}{dt} = \int (0.125 \times 10 \times E_i - \frac{dE_o}{dt}) dt$$

Figure 7.(b).

BLOCK VI - Power Piston

$$\frac{G6}{(1 + T8S)S}$$

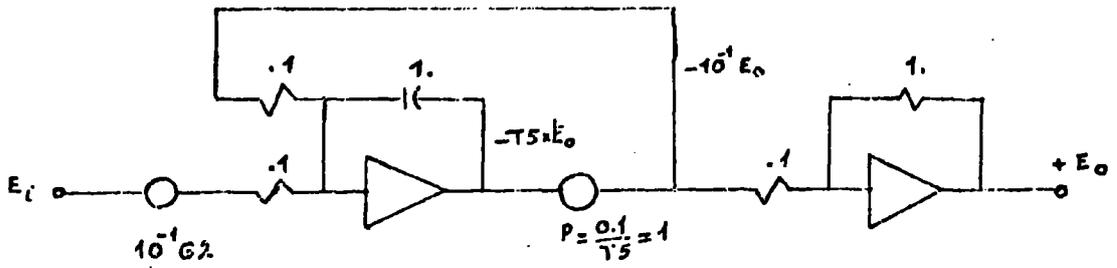
$$\frac{Eo}{Ei} = \frac{G6}{(1 + T8S)S}$$

$$T8 \frac{dEo}{dt} = \int (G6Ei - \frac{dEo}{dt}) dt$$

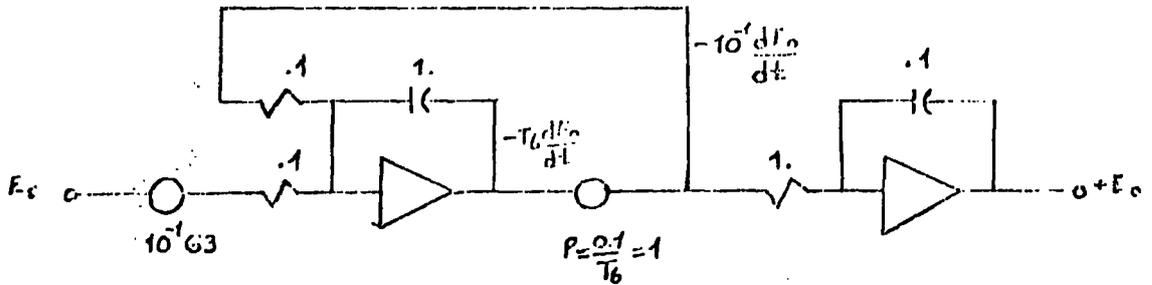
For $G6 = 0.0755$ and $T8 = 0.01$, it becomes

$$0.01 \frac{dEo}{dt} = \int (0.0755Ei - \frac{dEo}{dt}) dt$$

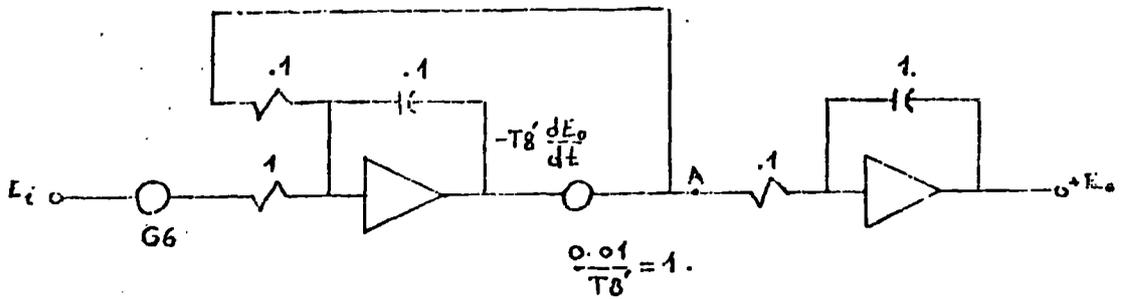
At the point A in the diagram , we have a signal corresponding to $10^{-1} \frac{dEo}{dt}$. See Figure 7. (c) overleaf.



(a)



(b)



(c)

Figure 7. Mathematical model of

(a) Servo-Valve

(b) Primary Actuator

(c) Power Piston

APPENDIX B

FIRMWARE DESCRIPTION AND PROGRAM LISTING

It was possible to punch the program onto computer cards and store the program in a memory file of the Newcastle IBM computer. Then this host computer could be instructed to compile the assembler coded program into machine code. An object tape of the machine code could then be obtained by using the departmental minicomputer Varian V73 because the Varian can easily be linked to the Newcastle computer.

But, our program was loaded via a keyboard and then the object tape of the program was obtained from the M6800. Since the software was developed step-by-step, entering the program parts via the keyboard was less time consuming than the procedure of obtaining the machine code by using the IBM computer and V73 microcomputer.

The software written for the project is given in the following pages. The program commands with their addresses in the successive memory locations are in a table form. The subroutines are arranged in two groups. GROUP-I includes the subroutines used to produce the output signals, read the input signal and store the data in the given memory locations. But the subroutines in GROUP-II are utilized to enable the TTY to print the commands to the operator.[15].



Test Program Listing

Memory Location	Object Code	Source Statements
0586	0F	SEI
0587	B6 800A	LDAA 800A
058A	86 30	LDAA \$E30
058C	B7 800B	STAA 800B
058F	4F	CLRA
0590	B7 800A *	STAA 800A
0593	01	NOP
.	.	.
.	.	.
05A6	01 *	NOP
05A7	B6 8008	LDAA 8008
05AA	B6 800C	LDAA 800C
05AD	86 32	LDAA \$E32
05AF	B7 8009	STAA 8009
05B2	B7 800D	STAA 800D
05B5	4F	CLRA
05B6	B7 8008	STAA 8008
05B9	B7 800C	STAA 800C
05BC	73 8008	COM 8008
05BF	73 800C *	COM 800C
05C2	86 0D LOOP	LDAA \$E0D
05C4	BD E1D1	JSR OUTEE
05C7	86 0A	LDAA \$E0A
05C9	BD E1D1	JSR OUTEE
05CC	86 28	LDAA \$E28
05CE	BD E1D1	JSR OUTEE

DUMMY CLEAR

IDENTIFY THE INPUT PORT

20 "No Operation" instructions

DUMMY CLEAR
DUMMY CLEAR

SELECT DDR1. CA2 TO LOW
SELECT DDR2. CA4 TO LOW

ESTABLISH D/A-1 AS OUTPUT
ESTABLISH D/A-2 AS OUTPUT

CARRIAGE RETURN

LINE FEED

PRINT:(1) PRESS
SELECT

APPLY "PEN ON" COMMAND

STAA 800D

IDENTIFY VALVE NUMBER

0637 B7 800D *

STAB 8008

063A F7 8008 LP2

LIX \$00FF

063D CE 00FF K3

DEX K3

0640 09 26 FD

BNE K3

0641 26 5C

DECB 0002

0643 F1 0002

CMPS LP2

0644 23 F1 *

BLS

NOP

0649 01

NOP

064A 01 *

LIX \$04FF

064B CE 04FF K4

DEX K4

064E 09 26 FD

BNE K4

064F 26 36

LDAA \$E36

0651 86 36

STAA 800D

0653 B7 800D *

STAB 8008

0656 F7 8008 LP3

LIX \$00FF

0659 CE 00FF K5

DEX K5

065C 09 26 FD

BNE K5

065D 26 5A

DECB \$E00

065F 5A C1 00

CMPS LP3

0660 C1 00 26 F2

BNE

0662 26 0001 *

LDAB 0001

0664 F6 0001

NOP

0667 01

NOP

0668 01

STAB 800C

0669 F7 800C LP4

LIX \$00FF

066C CE 00FF K6

DEX K6

066F 09 26 FD

BNE K6

0670 26 5A

DECB \$E00

0672 5A C1 00

CMPS

0673 C1 00

DELAY LOOP

RELEASE PEN

SET X TO ZERO

SET Y TO ZERO

0675	26	F2	BNE	LP4		
					*	
0677	01		NOP			7 "No Operation" instructions
.	.		.			
.	.		.			
067D	01		NOP			
					*	
067E	B7	8009	STAA	8009		SET CA2 TO HIGH
					*	
0681	01		NOP			5 "No Operation" instructions
.	.		.			
.	.		.			
0685	01		NOP			
					*	
0686	86	00	LDAA	\$E00		STORE SETPOINT
0688	B7	0F00	STAA	0F00		IN MEN-LOC OF00
068B	B7	800C	STAA	800C		APPLY IT TO MODEL
068E	CE	FFFF	LIX	\$EFFFF		
0691	09		DEX			
					K7	
0692	26	FD	BNE	K7		
0694	CE	FFFF	LIX	FFFF		
0697	09		DEX			
					K8	
0698	26	FD	BNE	K8		
					*	
069A	86	32	LDAA	\$E32		*(2)PRESS TEST PUSHBUTTON
069C	BD	E1D1	JSR	OUTIEE		
069F	BD	0314	JSR	PRESS		
06A2	BD	03CE	JSR	TEST		
06A5	BD	035A	JSR	PUSBUT		
06A8	B6	800E	LDAA	800E		DUMMY CLEAR NECESSARY
06AB	86	16	LDAA	\$E16		TO DO TEST AGAIN
06AD	B7	800F	STAA	800F		SET MODE CONTROL
06B0	B6	800F	LDAA	800F		
					LP5	
06B3	49		ROLA			
06B4	49		ROLA			

06B5	24 F9	BCC LP5	IS TEST ON?
06E7	01	NOP	11 "No Operation" instructions
06C1	01	NOP	
06C2	BD 08C5	JSR VALPOS	HEAD POSITION OUTPUT
06C5	01	NOP	STORE IN MEM-LOC 0503
06FE	01	NOP	58 "No Operation" instructions
06FF	86 3E	LDA \$E3E	
0701	B7 800D	STAA 800D	APPLY "PEN ON" COMMAND
0704	B6 0F00	LDA 0F00	NEGATIVE GOING
0707	BD 088F	JSR NGOING	PALP SIGNAL
070A	01	NOP	19 "No Operation" instructions
071C	01	NOP	
071D	86 33	LDA \$E33	
071F	BD E1D1	JSR OUTEE	PRINT: 3) PRESS TRIP
0722	BD 0314	JSR PRESS	FUSEBUTTON
0725	BD 033D	JSR TRIP	(
0728	BD 035A	JSR FUSEBT	
072B	01	NOP	
072C	01	NOP	
072D	01	NOP	
072E	B6 800C	LDA 800C	DULLY CLEAR
0731	86 3E	LDA \$E3E	
0733	B7 800D	STAA 800D	SET MODE CONTROL

*(3)PRESS TRIP FUSEBUTTON

0736	B6 800D	LDAA 800D			
0739	2A FE	BFL LP6		IS TRIP ON?	
073B	01	NOP		21 "No Operation" instructions	
074F	01	.			
		.			
		NOP			
* (4) RELEASE TEST PUSHBUTTON					
0750	86 34	LDAA \$E34		PRINT: 4) RELEASE TEST	
0752	BD E1 D1	JSR OUTEE		PUSHBUTTON	
0755	BD 0402	JSR RELEAS		(
0758	BD 03CE	JSR TEST			
075B	BD 035A	JSR PUSBUT			
075E	01	NOP			
075F	01	NOP			
0760	01	NOP			
0761	B6 800E	LDAA 800E		DUMMY CLEAR	
0764	86 04	LDAA \$E04			
0766	B7 800F	STAA 800F		SET MODE CONTROL	
0769	B6 800F	LDAA 800F			
076C	49	ROLA			
076D	49	ROLA			
076E	24 F9	BCC LP7		IS TEST OFF?	
0770	01	NOP		36 "No Operation" instructions	
		.			
		.			
0793	01	NOP			
0794	BD 08A6	JSR PGOING		CREATE POSITIVE GOING RAMP	
0797	01	NOP		28 "No Operation" instructions	
		.			
		.			
07B2	01	NOP			
* (5) PRESS TEST PUSHBUTTON					

07B3	86 35	LDAA	\$E35	PRINT: 5)PRESS TEST
07B5	BD E1D1	JSR	OUTEE	PUSHBUTTON
07B8	BD 0314	JSR	PRESS	(
07BB	BD 03CE	JSR	TEST	
07BE	BD 035A	JSR	PUSBUT	
07C1	01	NOP		
07C2	B6 800E	LDAA	800E	DUMMY CLEAR
07C5	86 16	LDAA	\$E16	
07C7	B7 800F	STAA	800F	SET MODE CONTROL
07CA	B6 800F	LDAA	800F	
07CD	49	ROLA		
07CE	49	RCLA		
07CF	24 F9	BCC	LP8	IS TEST ON?
07D1	01	NOP		
07D2	01	NOP		
07D3	17	TBA		
07D4	BD 088F	JSR	NGCING	CREATE NEGATIVE GOING RALIP
07D7	01	NOP		21 "No Operation" instructions
.	.	.		
.	.	.		
07EB	01	NOP		
*(6)RELEASE TRIP PUSHBUTTON				
07EC	86 36	LDAA	\$E36	PRINT: 6) RELEASE TRIP
07EE	BD E1D1	JSR	OUTEE	PUSHBUTTON
07F1	BD 0402	JSR	RELEAS	(
07F4	BD 033D	JSR	TRIP	
07F7	BD 035A	JSR	PUSBUT	
07FA	B6 800C	LDAA	800C	DUMMY CLEAR
07FD	86 3C	LDAA	\$E3C	
07FF	B7 800D	STAA	800D	SET MODE CONTROL
0802	B6 800D	LDAA	800D	
0805	2A FB	BPL	LP9	IS TRIP OFF?
0807	01	NOP		5 "No Operation" instructions
.	.	.		
.	.	.		
080B	01	NOP		

```

* (7) RELEASE TEST PUSHBUTTON
080C 86 37          LDA  $E37          PRINT: 7) RELEASE TEST
080E BD E1D1       JSR  OUTEE          PUSHBUTTON
0811 BD 0402       JSR  RELEAS         (
0814 BD 03CE       JSR  TEST
0817 BD 035A       JSR  PUSHBT
081A B6 800E       LDA  $00E          DUMMY CLEAR
081D 86 04         LDA  $E04          SET MODE CONTROL
081F B7 800F       STA  $00F          LPA
0822 B6 800F       LDA  $00F          LPA
0825 49           ROLA
0826 49           ROLA
0827 24 F9        ZCC  LPA          IS TEST OFF?
*
0829 01          NOP          14 "No Operation" instructions
.
.
0836 01          NOP
*
0837 B6 0503       LDA  0503          CREATE POSITIVE GOING RAMP
083A 5F          CLRB
083B BD 08A6       JSR  PGOING
*
083E 01          NOP          23 "No Operation" instructions
.
.
0854 01          NOP
*
* (8) RELEASE SELECT PUSHBUTTON
0855 86 38          LDA  $E38          PRINT: 8) RELEASE SELECT
0857 BD E1D1       JSR  OUTEE          PUSHBUTTON
085A BD 0402       JSR  RELEAS         (
085D BD 03E3       JSR  SELECT
0860 BD 035A       JSR  PUSHBT
0863 01          NOP          7 "No Operation" instructions
.
.
0869 01          NOP

```

086A	B6 8008	LDAA 8008	DUMMY CLEAR
086D	86 34	LDAA \$E34	
085F	B7 8009	STAA 8009	SET MODE CONTROL
0872	B6 8009	LDAA 8009	
0875	2A FB	BPL LPB	IS SELECT OFF?
0877	01	NOP	7 "No Operation" instructions
087D	01	NOP	
087E	86 36	LDAA \$E36	
0880	B7 800D	STAA 800D	RELEASE PER
0883	01	NOP	9 "No Operation" instructions
088B	01	NOP	
088C	7E 05C2	JMP LOOP	RUN PROG AGAIN
088F	B7 800C	NGOINGSTAA 800C	
0892	CE 1CB6	LIX \$E1CB6	NEGATIVE GOING ON D/A-1
0895	09	SDI	
0896	26 FD	BNE SDI	
0898	01	NOP	
0899	01	NOP	
089A	01	NOP	
089B	01	NOP	
089C	01	NOP	
089D	4A	DECA	
089E	26 EF	BNE NGOING	
08A0	39	RTS	
08A1	01	NCP	5 "No Operation" instructions
08A5	01	NOP	
08A6	F7 800C	PGOINGSTAB 800C	POSITIVE GOING ON D/A-1

08A9	CE 1CB6	LIX	\$E1CE6		
08AC	09	DEX			
08AD	26 FD	BNE	SD2		
08AF	01	NOP			
08B0	01	NOP			
08B1	01	NOP			
08B2	5C	INCB			
08B3	F1 OF00	CFEB	OF00		
08B6	23 EE	BLS	FGING		
08B8	39	RTS			
		*			
08B9	01	NOP			12 "No Operation" instructions
	.	.			
	.	.			
08C4	01	NOP			
		*			
		*	READ VALUE POSITION		
08C5	86 34	VALPOS	LDAA \$E34		
08C7	B7 800B	STAA	800B		
08CA	01	NOP			
08CB	86 3C	LDAA	\$E3C		
08CD	B7 800B	STAA	800B		CONVERT COMMAND TO HIGH A/D CONVERSION RUNS
08D0	B6 800B	LDAA	800B		
08D3	2A FB	BPL	SD3		
08D5	B6 800A	LDAA	800A		READ DATA IN
08D8	B7 0503	STAA	0503		SCORE DATA AT 0503
08DB	39	RTS			
		*			
08DC	01	NOP			10 "No Operation" instruction
	.	.			
	.	.			
08E5	01	NOP			
		*			
		*	X-Y VALUES FOR VALUE IDENTIFICATION		

08E6	86 3E	VALNUM	IDAA	\$E3E	CA2 TO HIGH
08E8	B7 8009		STAA	8009	
08EB	86 F5		IDAA	\$EF5	
08ED	B7 0000		STAA	0000	MINIMUM X
08F0	86 9C		IDAA	\$E9C	
08F2	B7 0001		STAA	0001	Y
08F5	86 FE		IDAA	\$EFE	
08F7	B7 0002		STAA	0002	MAXIMUM X
08FA	39		RTS		

TTY Printing Program Listing

Memory Location Object Code Source Statements

```

* PRINT : ) PRESS
0314 86 29 PRESS LDA $E29
0316 BD E1D1 JSR OUTEE
0319 86 20 LDA $E20
031B BD E1D1 JSR OUTEE
031E 86 50 LDA $E50
0320 BD E1D1 JSR OUTEE
0323 86 52 LDA $E52
0325 BD E1D1 JSR OUTEE
0328 86 45 LDA $E45
032A BD E1D1 JSR OUTEE
032D 86 53 LDA $E53
032F BD E1D1 JSR OUTEE
0332 86 53 LDA $E53
0334 BD E1D1 JSR OUTEE
0337 86 20 LDA $E20
0339 BD E1D1 JSR OUTEE
033C 39 RTS
SPACE
    
```

```

* PRINT : TRIP
033D 86 54 TRIP LDA $E54
033F BD E1D1 JSR OUTEE
0342 86 52 LDA $E52
0344 BD E1D1 JSR OUTEE
0347 86 49 LDA $E49
0349 BD E1D1 JSR OUTEE
034C 86 50 LDA $E50
034E BD E1D1 JSR OUTEE
0351 39 RTS
    
```

8 "No Operation" instructions

0352	01	NOP	
.	.	.	
.	.	NOP	
0359	01		
* PRINT : PUSHBUTTON			
035A	86 20	FUSBUT	LDAA \$C20
035C	BD E1D1	JSR	OUTEE
035F	86 50	LDAA	\$C50
0361	BD E1D1	JSR	OUTEE
0364	86 55	LDAA	\$C55
0366	BD E1D1	JSR	OUTEE
0369	86 53	LDAA	\$C53
036B	BD E1D1	JSR	OUTEE
036E	86 48	LDAA	\$C48
0370	BD E1D1	JSR	OUTEE
0373	86 42	LDAA	\$C42
0375	BD E1D1	JSR	OUTEE
0378	86 55	LDAA	\$C55
037A	BD E1D1	JSR	OUTEE
037D	86 54	LDAA	\$C54
037F	BD E1D1	JSR	OUTEE
0382	86 54	LDAA	\$C54
0384	BD E1D1	JSR	OUTEE
0387	86 4F	LDAA	\$C4F
0389	BD E1D1	JSR	OUTEE
038C	86 4E	LDAA	\$C4E
038E	BD E1D1	JSR	OUTEE
0391	86 0D	LDAA	\$C0D
0393	BD E1D1	JSR	OUTEE
0396	86 0A	LDAA	\$C0A
0398	BD E1D1	JSR	OUTEE
039B	86 28	LDAA	\$C28
039D	BD E1D1	JSR	OUTEE
03A0	39	RTS	

SPACE

CARRIAGE RETURN

LINE FEED

45 "No Operation" Instructions

```

03A1 01 NOP
      .
      .
03CD 01 NOP
      .
      * PRINT : TEST
      TEST LDA $E54
           JSR OUTEE
           LDA $E45
           JSR OUTEE
           LDA $E53
           JSR OUTEE
           LDA $E54
           JSR OUTEE
           RTS
03E2 39
03CE 86 54
03D0 BD E1D1
03D3 86 45
03D5 BD E1D1
03D8 86 53
03DA BD E1D1
03DD 86 54
03DF BD E1D1
03E2 39

```

```

03E3 86 53
03E5 BD E1D1
03E8 86 45
03EA BD E1D1
03ED 86 4C
03EF BD E1D1
03F2 86 45
03F4 BD E1D1
03F7 86 43
03F9 BD E1D1
03FC 86 54
03FE BD E1D1
0401 39

```

```

      * PRINT : SELECT
      SELECT LDA $E53
            JSR OUTEE
            LDA $E45
            JSR OUTEE
            LDA $E4C
            JSR OUTEE
            LDA $E45
            JSR OUTEE
            LDA $E43
            JSR OUTEE
            LDA $E54
            JSR OUTEE
            RTS
      * PRINT : RELEASE
      RELEAS LDA $C29
            JSR OUTEE
            LDA $C20
            JSR OUTEE

```

SPACE

040C	86	52	LDAA	\$C52
040E	BD	E1D1	JSR	OUTEE
0411	86	45	LDAA	\$E45
0413	BD	E1D1	JSR	OUTEE
0416	86	4C	LDAA	\$E4C
0418	BD	E1D1	JSR	OUTEE
041B	86	45	LDAA	\$E45
041D	BD	E1D1	JSR	OUTEE
0420	86	41	LDAA	\$E41
0422	BD	E1D1	JSR	OUTEE
0425	86	53	LDAA	\$E53
0427	BD	E1D1	JSR	OUTEE
042A	86	45	LDAA	\$E45
042C	BD	E1D1	JSR	OUTEE
042F	86	20	LDAA	\$E20
0431	BD	E1D1	JSR	OUTEE
0434		39	RTS	

SPACE

Machine Code Listing of the TTY Printing Program

41. 0

S11303148620BDE1D18620BDE1D18650BDE1D18607
S1130324520BDE1D18645BDE1D18653BDE1D18653A9
S113033460BDE1D18620BDE1D186554BDE1D1865207
S113034480L1D18648BDE1D18650BDE1D13901017F
S11303540101010101018620BDE1D18650BDE1D135
S11303648655BDE1D18653BDE1D18648BDE1D18630
S113037442BDE1D18655BDE1D18654BDE1D1865457
S11303843BDE1D1864FBDE1D1864EBDE1D18605B1F
S1130394E1D1860ABDE1D18628BDE1D1390101014B
S11303A40101010101010101010101010101010135
S11303B49101010101010101010101010101010125
S11303C491010101010101010101018654BDE1D1864C
S11303D445BDE1D18653BDE1D18654BDE1D1398511
S11303E453BDE1D18645BDE1D1864CBDE1D18645FD
S11303F48BDE1D18643BDE1D18654BDE1D139862910
S11304043BDE1D18620BDE1D18650BDE1D186453B91
S1130414E1D1864CBDE1D18645BDE1D18641BDE142
S1130424D18653BDE1D18645BDE1D18620BDE1D15C
C1050434390039

Memory Locations		The Purposes of the Program Instructions
From	To	
0586	-	Inhibiting the microprocessor from servicing an interrupt from a peripheral device. This instruction disables the MIKBUG interrupt routine and enables the use of the interrupt request flags of the PIA control registers as test bits without causing the program to jump into a service interrupt routine.
0587	05C1	Supervising the whole program in general terms. But the instructions in the locations 05AD to 05B4 of this section identify CA2 and CA4 as outputs, and make CA2 and CA4 staying at low.
05C2	05DE	Enabling the teletype to print: (1) PRESS SELECT PUSHBUTTON (
05E5	05F1	Checking if the SELECT is on or not.
061A	0634	Supplying the defined values to the X input and the Y input of the plotter. These values are hexa-decimal F5 for the X axis and 9C for the Y axis. There are two delay loops, one of which is after the X input is given (locations 0623 to 0628) and the other is after the Y input is given (locations 062F to 0634). These delay loops are necessary to obtain the smooth jumping of the pen of the plotter to the valve number on the plotter

paper. The Y axis value, the minimum and the maximum X axis values are stored in the memory locations 0001, 0000 and 0002 respectively via the VALNUM subroutine. As the SELECT command reaches the MPU loads these values in their locations and then the starting values of the both axes are placed on the corresponding plotter axes.

- 0635 0639 Applying the PEN ON command to the plotter .
This is achieved by setting the PIA's control line CA4 to the logic 1 that corresponds to 5 volts.
- 063A 0648 Drawing an identifying line under the valve number. Since the number has already been reached by the plotter pen and the PEN ON command has been applied, increasing the value on the X axis to a constant value makes the pen draw a line parallel to the X axis of the plotter.
For this particular program, after the PEN ON has been applied, the computer output to the plotter's X input increases from the hexadecimal F5 to FF.
- 064B 0650 The delay loop. This delay loop ensures the smooth return to the set-point on the plotter.
- 0651 0655 Releasing the pen. The pen is released when CA4 is de-energised.

0656	0676	Via these instructions the pen moves to its set point. This is done by decreasing the values on the X and Y axis to zero in sequence. The delay loops identified by L4 and L5 are used to prevent the pen from sudden jumps.
067E	0680	Short-circuiting the relay R2 and switching the point B of the relay R3 to the point 2 by energising the PIA control line CA2.
0686	068D	Storing an input value to the mathematical model in the memory location OF00 and also giving the same value to the model itself. The value in the program is hexadecimal C0.
068E	0699	Two delay loops. Since the model itself takes 1.75 sec to response to the step input, these two loops are necessary to give enough time to the model to response adequately. Here two delay loops takes 2.1 seconds of execution time. The reason why 1.75 sec was not used was the fact that the plotter response was taking longer than that of the model.
069A	06A7	Enabling the TTY to print: 2) PRESS TEST PUSHBUTTON (

06A8	06B6	Checking if the TEST is on or not.
06C2	06C4	Reading the valve position via the VALPOS sub-routine.
06FF	0703	Application of the PEN ON command by setting the PIA control line CA4 to the logic 1.
0704	0709	Decreasing the model input to zero via the sub-routine NGOING.
071D	072A	Enabling the TTY to print: 3) PRESS TRIP PUSHBUTTON (
072E	073A	Checking if the TRIP is on.
0750	075D	Enabling the TTY to print: 4) RELEASE TEST PUSHBUTTON (
0761	0793	Checking if the TEST button has been released.
0794	0796	Enabling the computer to give a positive going signal to the model via the subroutine PGOING.

07B3	07C0	Enabling the TTY to print: 5) PRESS TEST PUSHBUTTON (
07C2	07D0	Checking if the TEST button has been pressed down.
07D3	07D6	Reducing the model input to zero through the subroutine NGOING.
07EC	07F9	Enabling the TTY to print: 6) RELEASE TRIP PUSHBUTTON (
07FA	0806	Checking if the TRIP button has been released
080C	0819	Enabling the TTY to print: 7) RELEASE TEST PUSHBUTTON (
081A	0828	Testing if the TEST button has been released.
083A	083D	Supplying back the operating input signal of the valve control device before the test sequencing had started.
0855	0862	Enabling the TTY to print: 8) RELEASE SELECT PUSHBUTTON (

086A	0876	Testing if the SELECT button has been released.
087E	0882	Releasing the PEN ON command.
088C	088E	Enabling the operator to do the test sequencing again.

GROUP I

Memory Locations		S u b r o u t i n e s	
From	To	Name	Purposes
088F	08A0	INGOING	Storing the contents of the accumulator A at the output address 800C, decreasing A by one and continuing the storing until the value in A becomes zero. This procedure is done so that 3 seconds of processing time are spent until the contents of A drops to zero from the hexadecimal FF.
08A6	08B9	PGOING	Storing the contents of the accumulator B at the output address 800C, incrementing B by one and comparing the new value with the contents of the memory location OF00. Storing continues until the value in B becomes equal to the value compared with. In this subroutine, increasing the contents of B from 00 to hexadecimal FF takes 3 sec of execution time.
08C5	08DB	VALPOS	Reading the model output at the input port address 800A and storing it in the memory location 0503.

08E6	08FA	VALNUM	
			Setting CA2 to high resulting in the fact that the relay R2 becomes open-circuit and the points 0 and 1 of the relay R3 becomes short-circuit.

			Storing the hexadecimal data F5, 9C and FE in the memory locations 0000, 0001 and 0002 respectively.
--	--	--	--

GROUP II

Memory Locations		S u b r o u t i n e s	
From	To	Name	Purposes
0314	033C	PRESS	Enabling the TTY to print a close-bracket and leave a space. Then the TTY prints the letters P, R, E, S and S. These printed letters are followed by a second space.
033D	0351	TRIP	Enabling the TTY to print T, R, I and P.
035A	03A0	PUSBUF	Leaving a space and printing the letters P, U, S, I, B, U, T, I, O and N. This operation is followed by a Carriage Return and a Line Feed command. Then the TTY prints an open-bracket character at the beginning of the next line.
03CE	03E2	TEST	The TTY prints the letters T, E, S and T.
03E3	0401	SELECT	The letters S, E, L, E, C and T are printed.
0402	0435	RELEAS	Printing a close-parenthesis character, leaving a space and then printing the letters R, E, L, E, A, S and E. A

second space is obtained after these
letters have been printed.

APPENDIX C

SELECTION OF A VALVE NUMBER ON THE PLOTTER PAPER USED WITH THE MICROPROCESSOR-BASED VALVE TESTING SYSTEM.

If it is desired to underline the number 14 on the plotter paper, the 'Lead Accumulator A' instructions are to be loaded with 245, 254 and 156 decimal numbers in the memory locations 08EB, 08F5, and 08F0 respectively. Then the specified setpoint is arranged. When the program runs, the plotter draws the line under 14 if the command SELECT is on.

Checking back can be done as the following:

Firstly, because of the specification of the conditional jump in the memory location 0647, during underlining the valve number 14 the maximum number on the Output Register that is connected to the converter D/A -1 on the X input path of the plotter is decimal 255 but not 254. This full scale reading gives 3.8 volts. The decimal 245 and 156 correspond to 3.65V and 2.32V respectively. Since the gain of the amplifier on the X-axis is 3.35, the signals 3.65V and 3.8V should be multiplied by 3.35. The gain of the amplifier on the Y-axis is 2.26. Then the signal 2.32V should be multiplied by 2.26.

$$3.65V \times 3.35 = 12.23V$$

$$3.8V \times 3.35 = 12.73V$$

$$2.32V \times 2.26 = 5.24 V$$

Since the scaling on both axes is the same 500mV/cm, the amplifiers' outputs should be divided by 0.5V/cm. Then the minimum

$$\text{value on the X-axis is } \frac{12.23V}{0.5V/cm} = 24.46 \text{ cm}$$

the maximum value on the X-axis is $\frac{12.73V}{0.5V/cm} = 25.46 \text{ cm}$

the value on the Y-axis is $\frac{5.24V}{0.5V/cm} = 10.5 \text{ cm}$

These are the distances from the setpoint of the plotter. As it will be seen from Figure 5f, these are the correct distances for the number 14.

